

# COMPLEX WAVELETS: WHAT ARE THEY AND WHAT CAN THEY DO?

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## COMPLEX WAVELETS: WHAT ARE THEY AND WHAT CAN THEY DO?

- Basic form of the DT CWT
- Shift invariance of subband transfer functions
- DT CWT in 2-D – directional selectivity
- DT CWT in 3-D
- Denoising
- Image Registration
- Accumulated maps for keypoint detection
- Rotation-invariant local feature matching

## FEATURES OF THE (REAL) DISCRETE WAVELET TRANSFORM (DWT)

- **Good compression** of signal energy.
- **Perfect reconstruction** with short support filters.
- **No redundancy**.
- **Very low computation** – order- $N$  only.

But

- **Severe shift dependence**.
- **Poor directional selectivity** in 2-D, 3-D etc.

The DWT is normally implemented with a tree of highpass and lowpass filters, separated by  $2 : 1$  decimators.

# REAL DISCRETE WAVELET TRANSFORM (DWT) IN 1-D

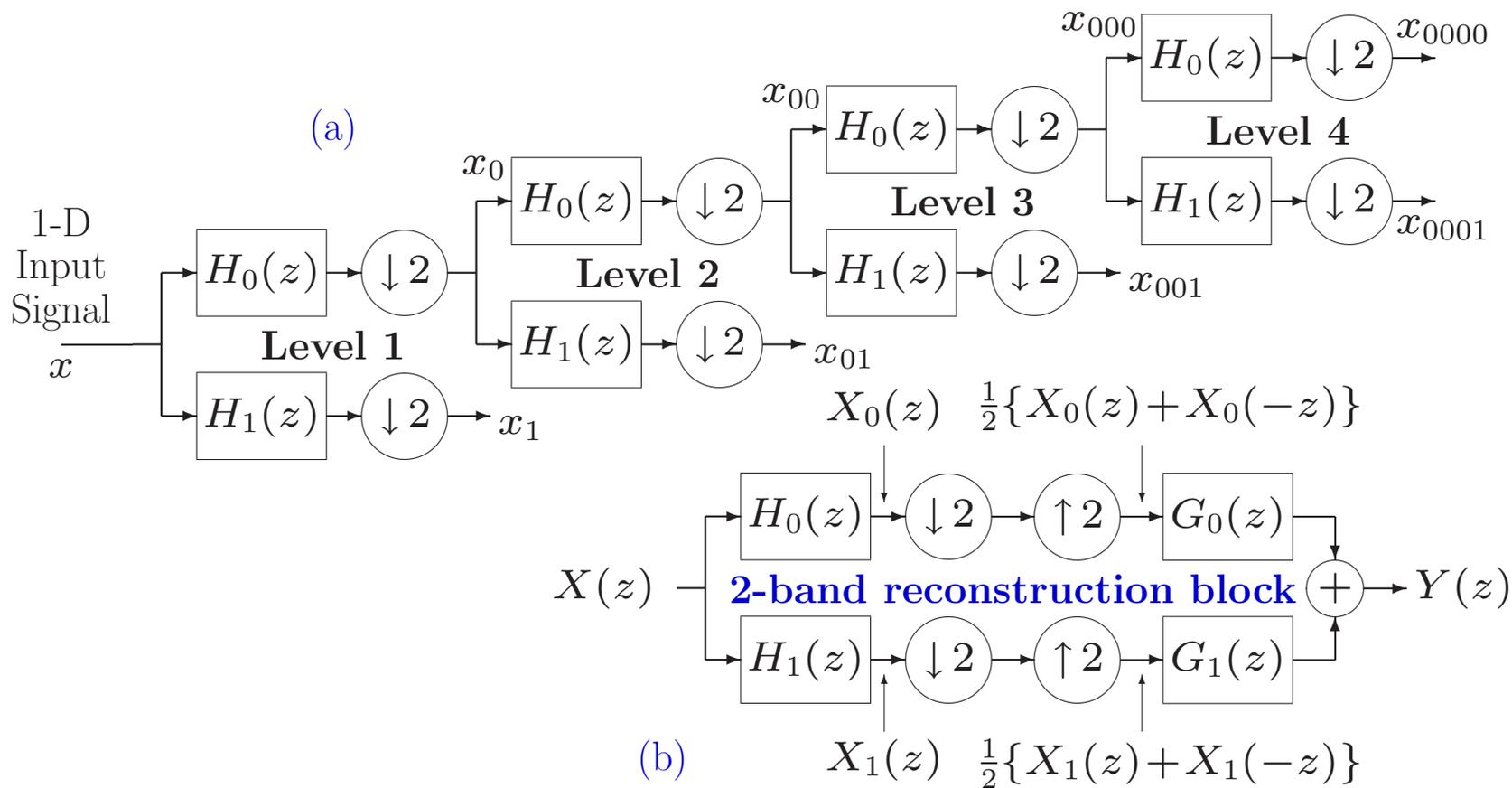


Figure 1: (a) Tree of real filters for the DWT. (b) Reconstruction filter block for 2 bands at a time, used in the inverse transform.

## VISUALISING SHIFT INVARIANCE

- Apply a standard input (e.g. unit step) to the transform for a **range of shift positions**.
- Select the transform coefficients from **just one wavelet level** at a time.
- Inverse transform each set of selected coefficients.
- Plot the component of the reconstructed output for each shift position at each wavelet level.
- Check for **shift invariance** (similarity of waveforms).

See Matlab demonstration.

## FEATURES OF THE DUAL TREE COMPLEX WAVELET TRANSFORM (DT CWT)

- Good **shift invariance** = **negligible aliasing**. Hence transfer function through each subband is independent of shift **and** wavelet coefs can be interpolated within each subband, independent of all other subbands.
- Good **directional selectivity** in 2-D, 3-D etc. – derives from **analyticity** in 1-D (ability to separate positive from negative frequencies).
- **Perfect reconstruction** with short support filters.
- **Limited redundancy** – 2:1 in 1-D, 4:1 in 2-D etc.
- **Low computation** – much less than the undecimated (à trous) DWT.

Each tree contains purely real filters, but the two trees produce the **real and imaginary parts** respectively of each complex wavelet coefficient.

# Q-SHIFT DUAL TREE COMPLEX WAVELET TRANSFORM IN 1-D

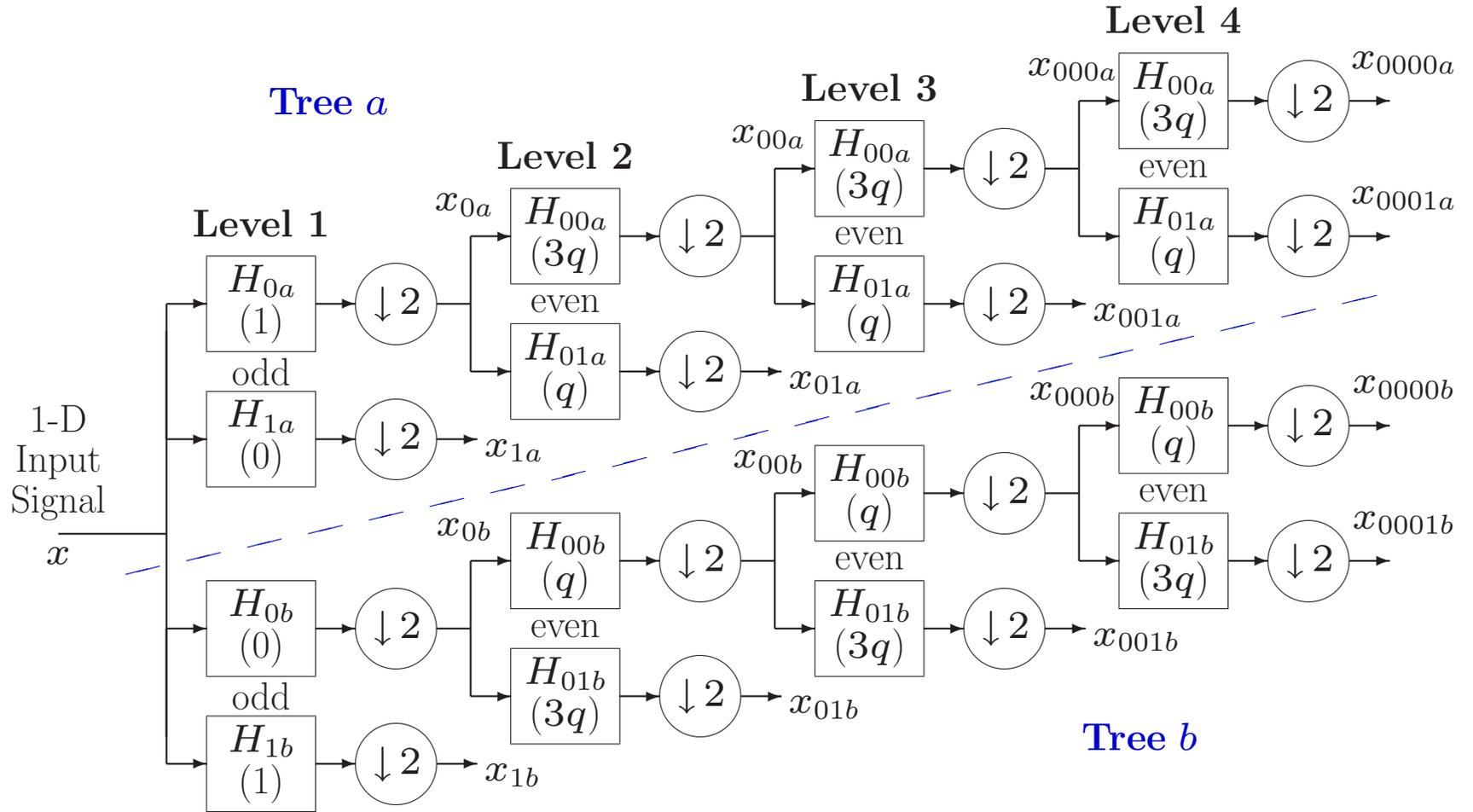


Figure 2: Dual tree of real filters for the Q-shift CWT, giving real and imaginary parts of complex coefficients from tree *a* and tree *b* respectively. Figures in brackets indicate the approximate delay for each filter, where  $q = \frac{1}{4}$  sample period.

## FEATURES OF THE Q-SHIFT FILTERS

Below level 1:

- Half-sample delay difference is obtained with filter delays of  $\frac{1}{4}$  and  $\frac{3}{4}$  of a sample period (instead of 0 and  $\frac{1}{2}$  a sample for our original DT CWT).
- This is achieved with an **asymmetric even-length** filter  $H(z)$  and its time reverse  $H(z^{-1})$ .
- Due to the asymmetry (like Daubechies filters), these may be designed to give an **orthonormal perfect reconstruction** wavelet transform.
- Tree **b** filters are the **reverse** of tree **a** filters, and reconstruction filters are the reverse of analysis filters, so **all filters** are from the **same orthonormal set**.
- Both trees have the **same frequency responses**.
- The combined **complex** impulse responses are **conjugate symmetric** about their mid points, even though the separate responses are asymmetric. Hence **symmetric extension** still works at image edges.

## Q-SHIFT DT CWT BASIS FUNCTIONS – LEVELS 1 TO 3

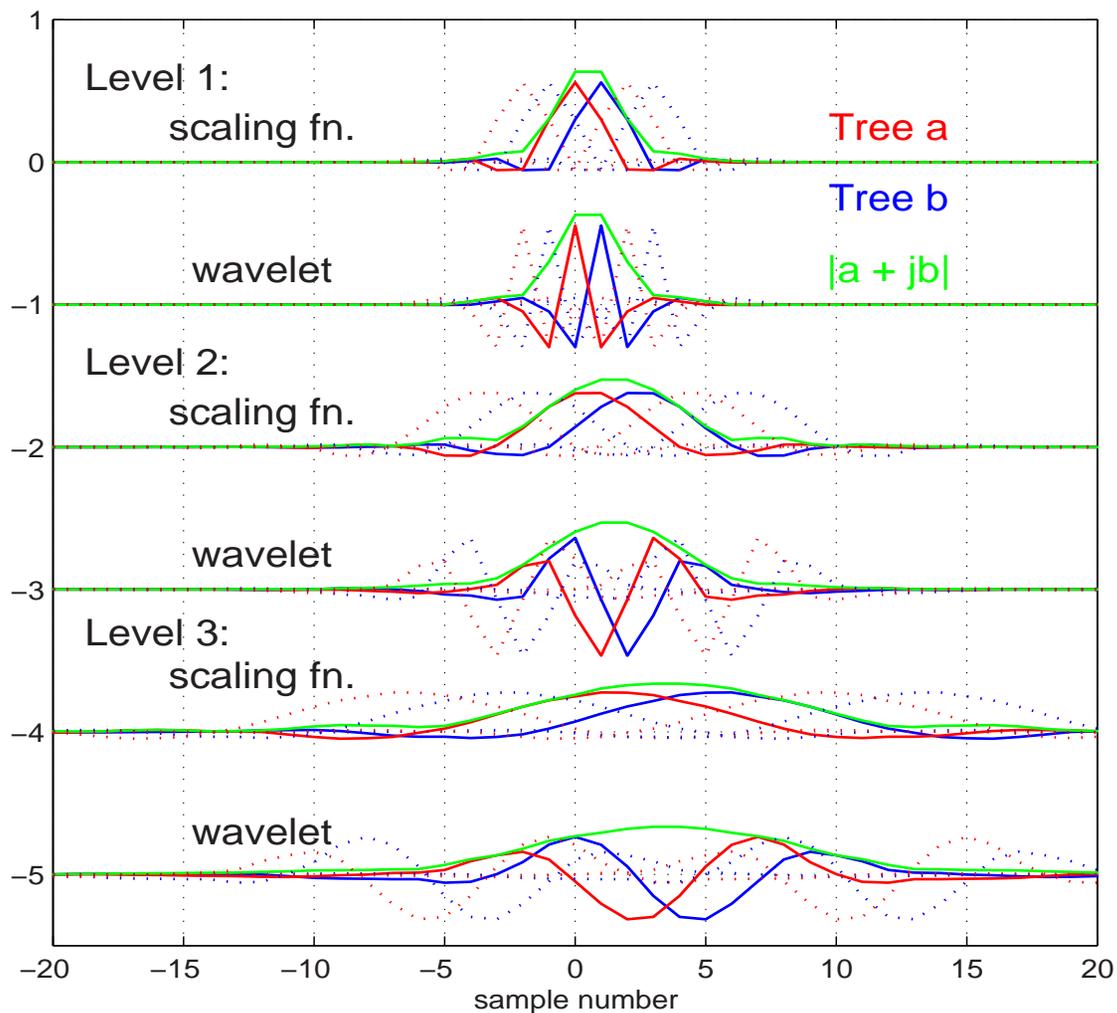
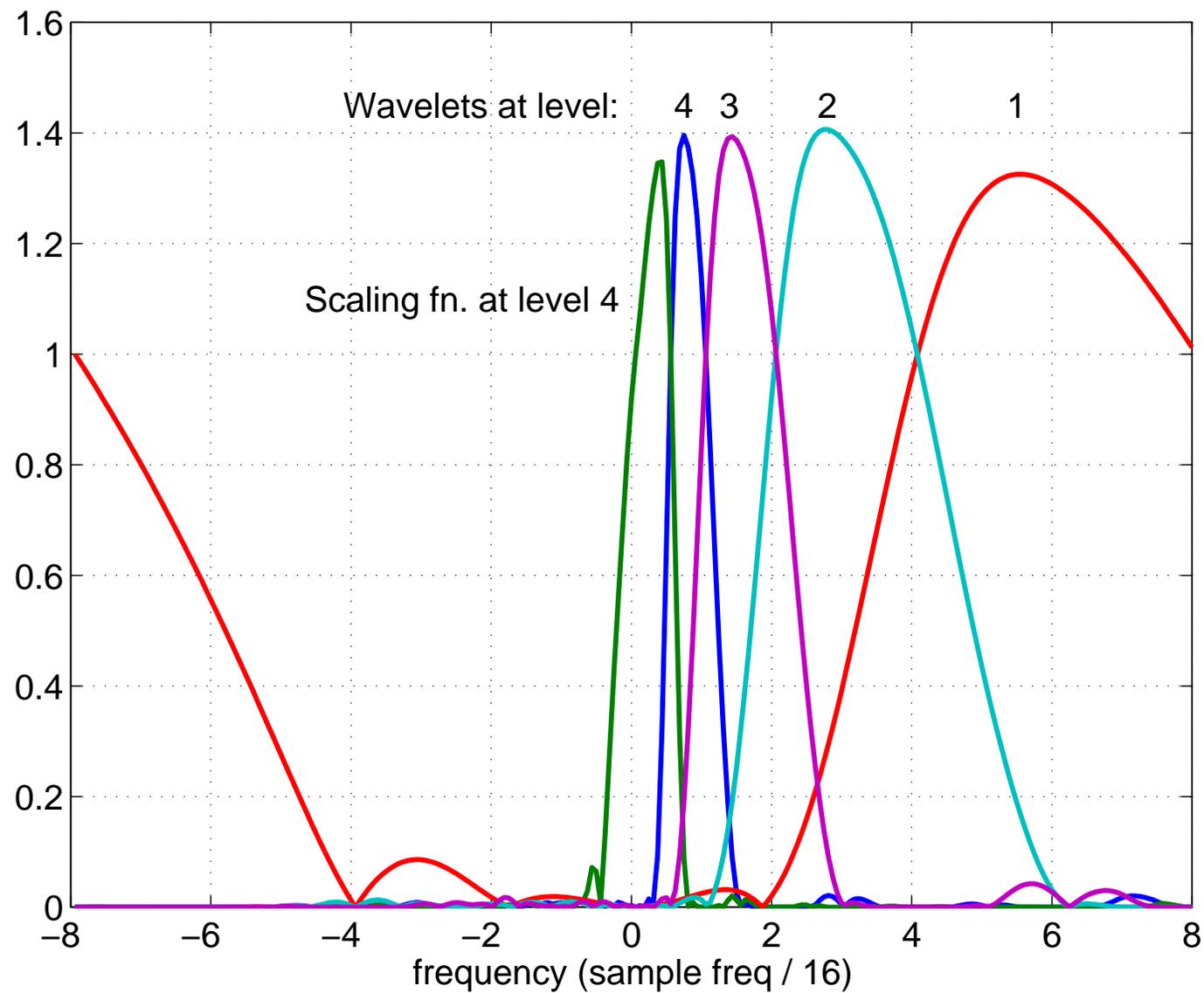
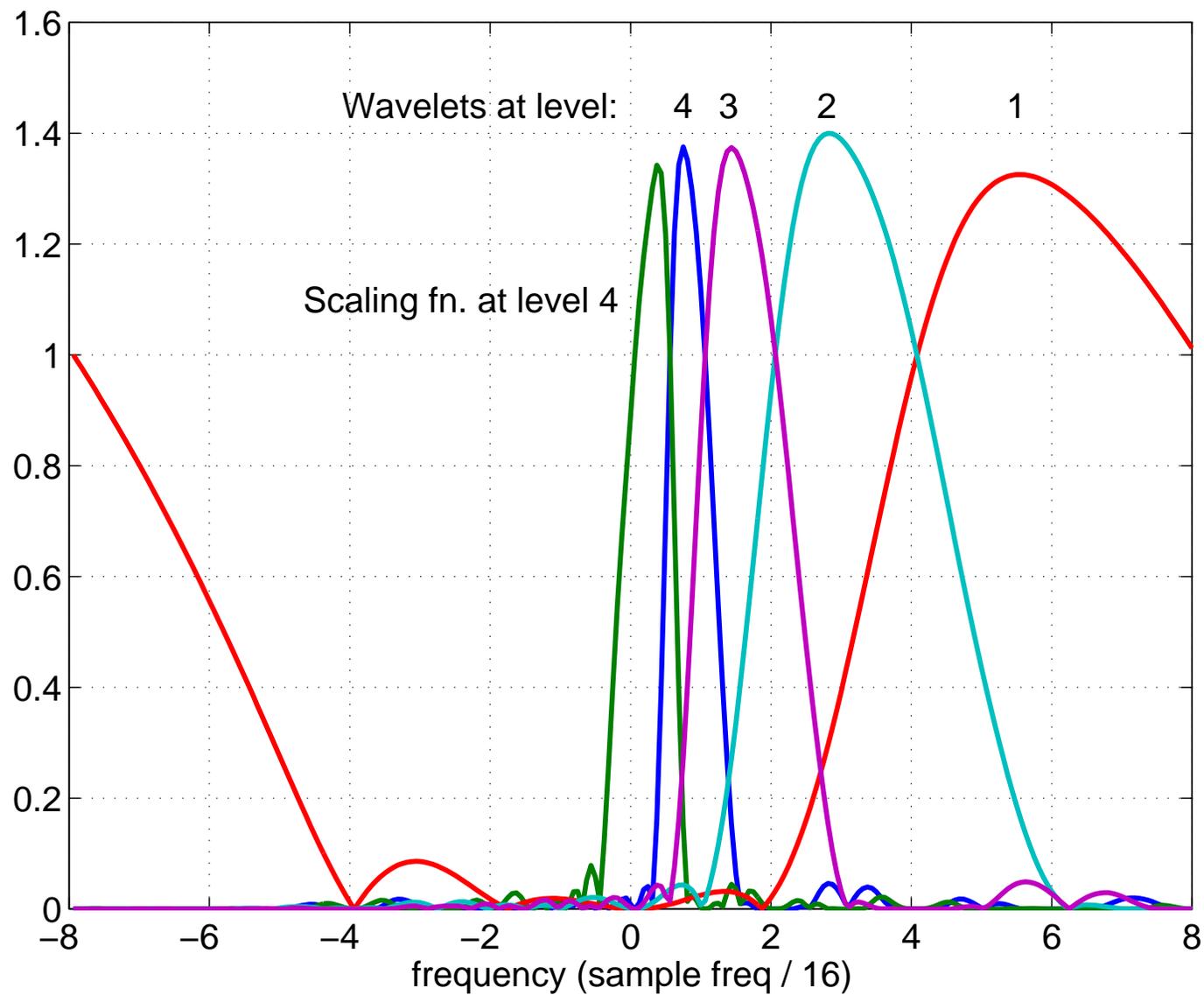


Figure 3: Basis functions for adjacent sampling points are shown dotted.

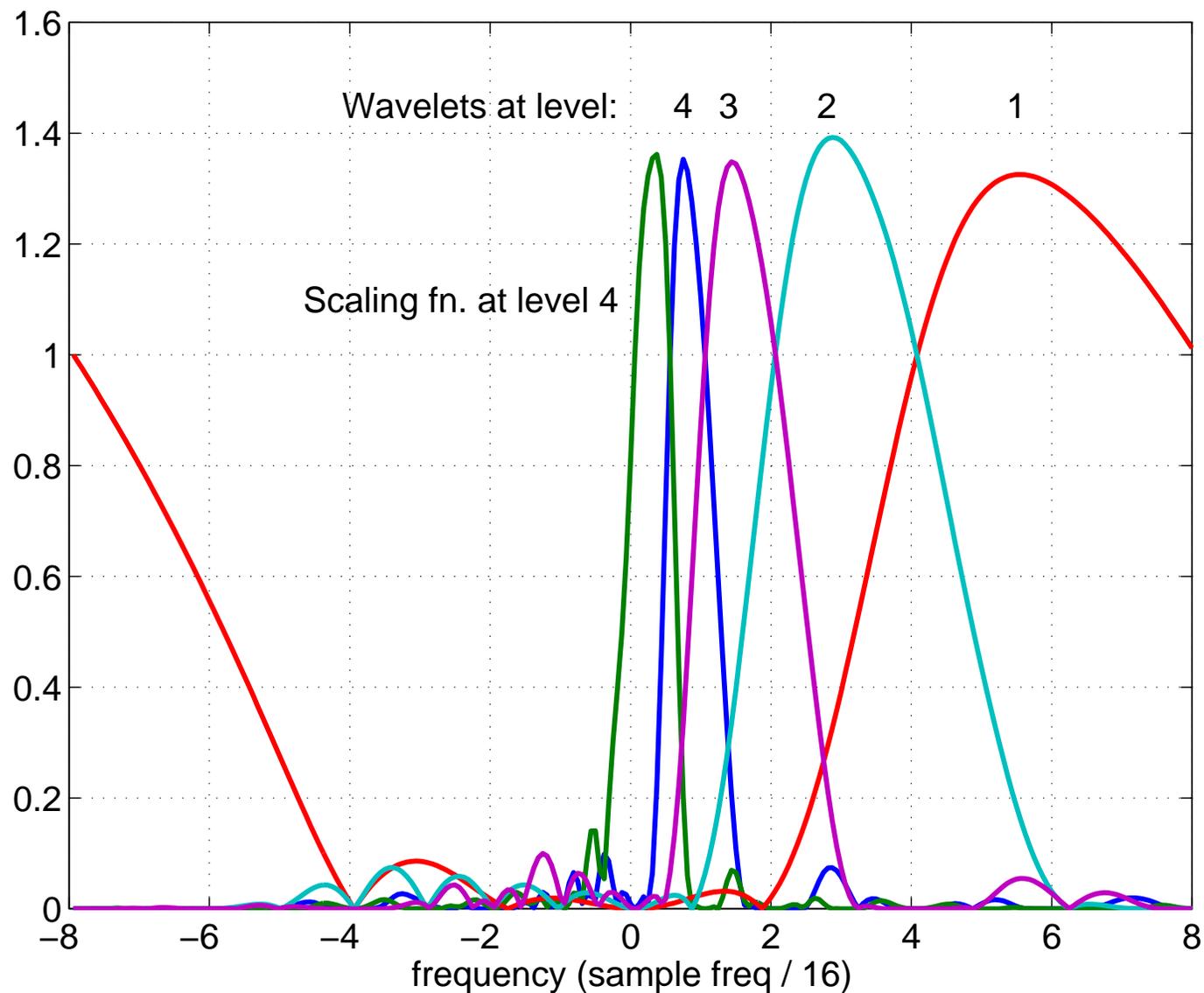
## FREQUENCY RESPONSES OF 18-TAP Q-SHIFT FILTERS



## FREQUENCY RESPONSES OF 14-TAP Q-SHIFT FILTERS



## FREQUENCY RESPONSES OF 6-TAP Q-SHIFT FILTERS

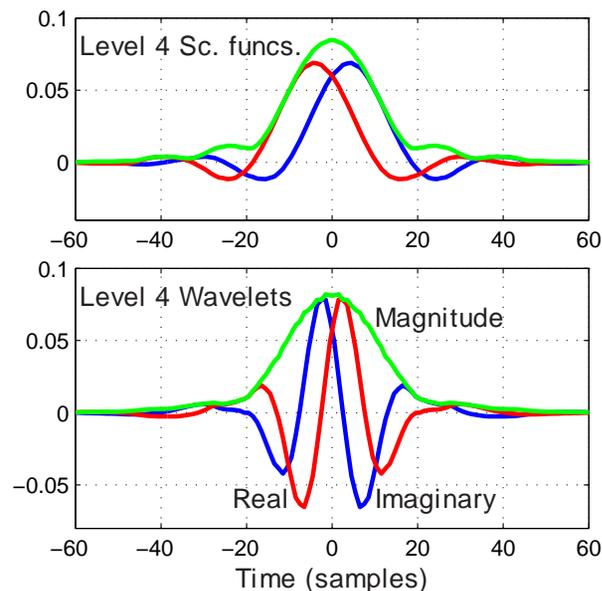
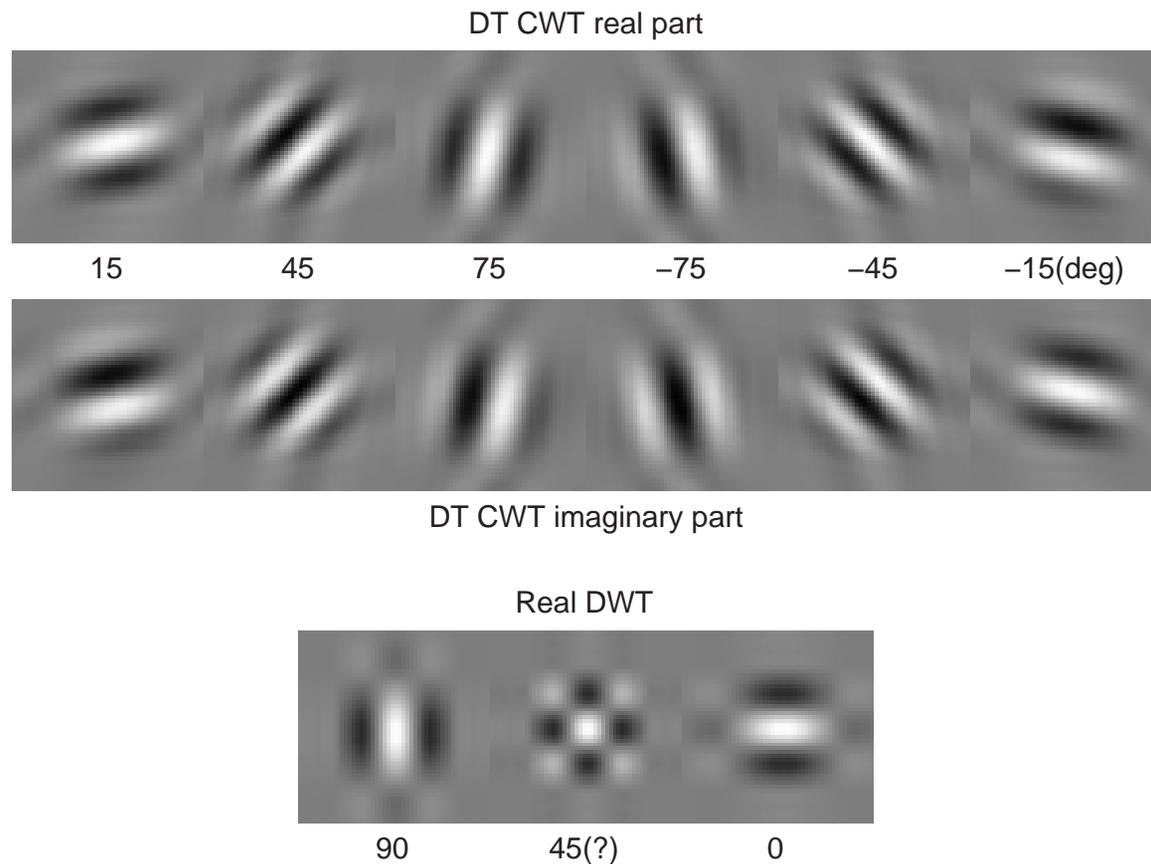


## THE DT CWT IN 2-D

When the DT CWT is applied to 2-D signals (images), it has the following features:

- It is performed separably, with 2 trees used for the rows of the image and 2 trees for the columns – yielding a **Quad-Tree** structure (4:1 redundancy).
- The 4 quad-tree components of each coefficient are combined by simple sum and difference operations to yield a **pair of complex coefficients**. These are part of two separate subbands in adjacent quadrants of the 2-D spectrum.
- This produces **6 directionally selective subbands** at each level of the 2-D DT CWT. Fig 4 shows the basis functions of these subbands at level 4, and compares them with the 3 subbands of a 2-D DWT.
- The DT CWT is directionally selective (see fig 6) because the complex filters can **separate positive and negative frequency components** in 1-D, and hence **separate adjacent quadrants** of the 2-D spectrum. Real separable filters cannot do this!

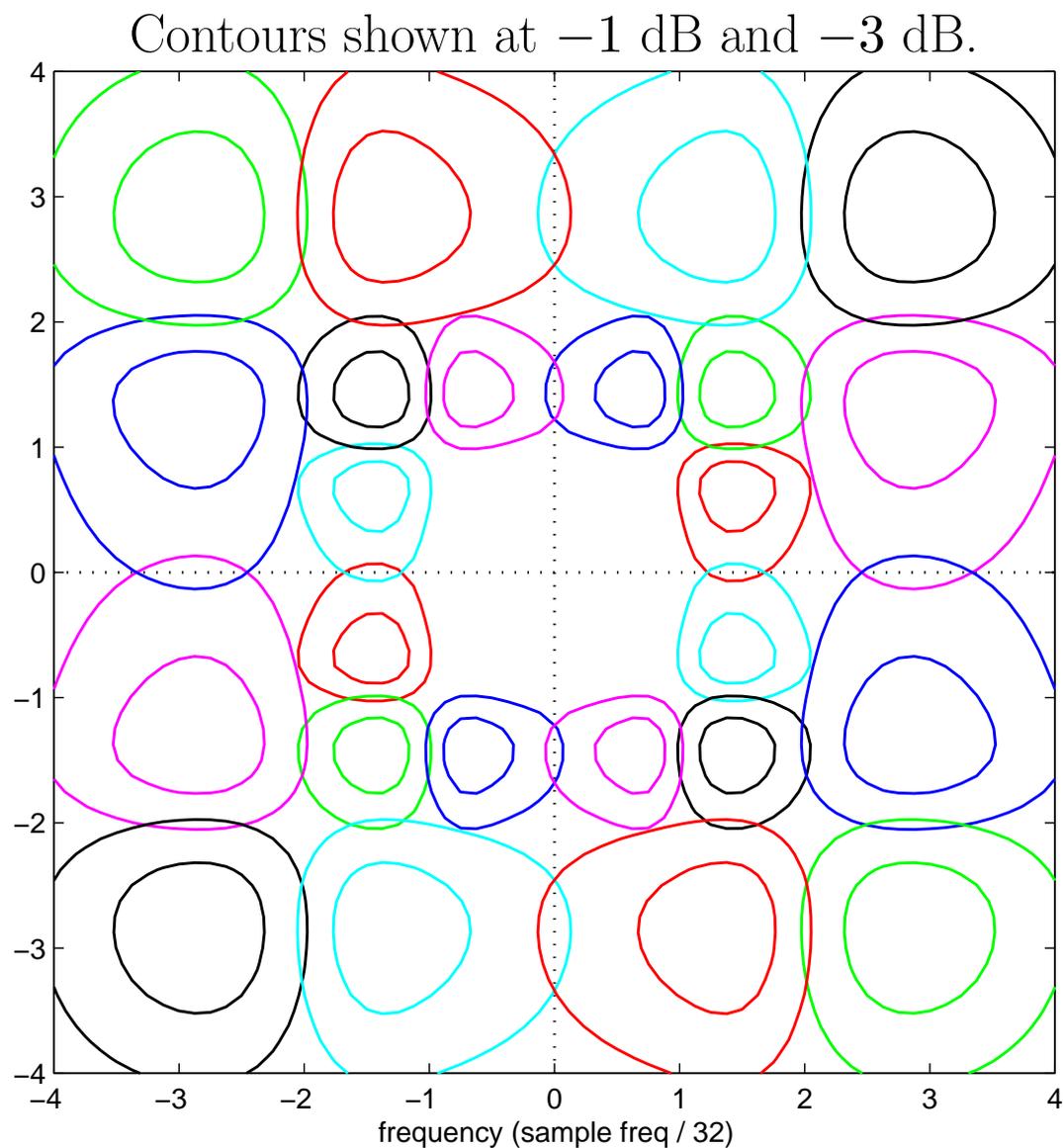
## 2-D BASIS FUNCTIONS AT LEVEL 4



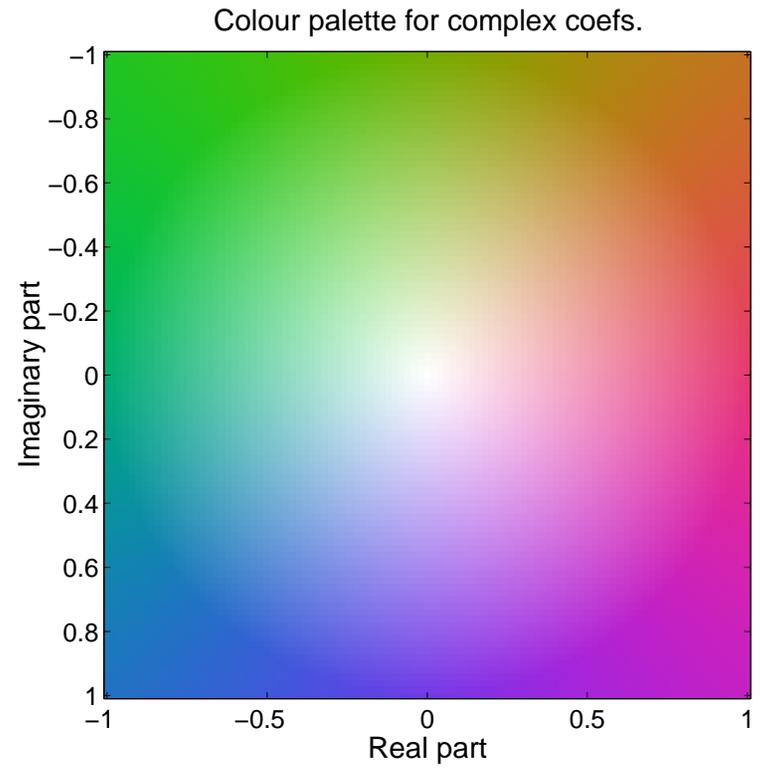
$$e^{j\omega_1 x} e^{j\omega_2 y} = e^{j(\omega_1 x + \omega_2 y)}$$

Figure 4: Basis functions of 2-D Q-shift complex wavelets (top), and of 2-D real wavelet filters (bottom), all illustrated at level 4 of the transforms. The complex wavelets provide 6 directionally selective filters, while real wavelets provide 3 filters, only two of which have a dominant direction. The 1-D bases, from which the 2-D complex bases are derived, are shown to the right.

## FREQUENCY RESPONSES OF 2-D Q-SHIFT FILTERS AT LEVELS 3 AND 4



## TEST IMAGE AND COLOUR PALETTE FOR COMPLEX COEFFICIENTS



## 2-D DT-CWT DECOMPOSITION INTO SUBBANDS



Figure 5: Four-level DT-CWT decomposition of *Lena* into 6 subbands per level (only the central  $128 \times 128$  portion of the image is shown for clarity). A colour-wheel palette is used to display the complex wavelet coefficients.

## 2-D DT-CWT RECONSTRUCTION COMPONENTS FROM EACH SUBBAND

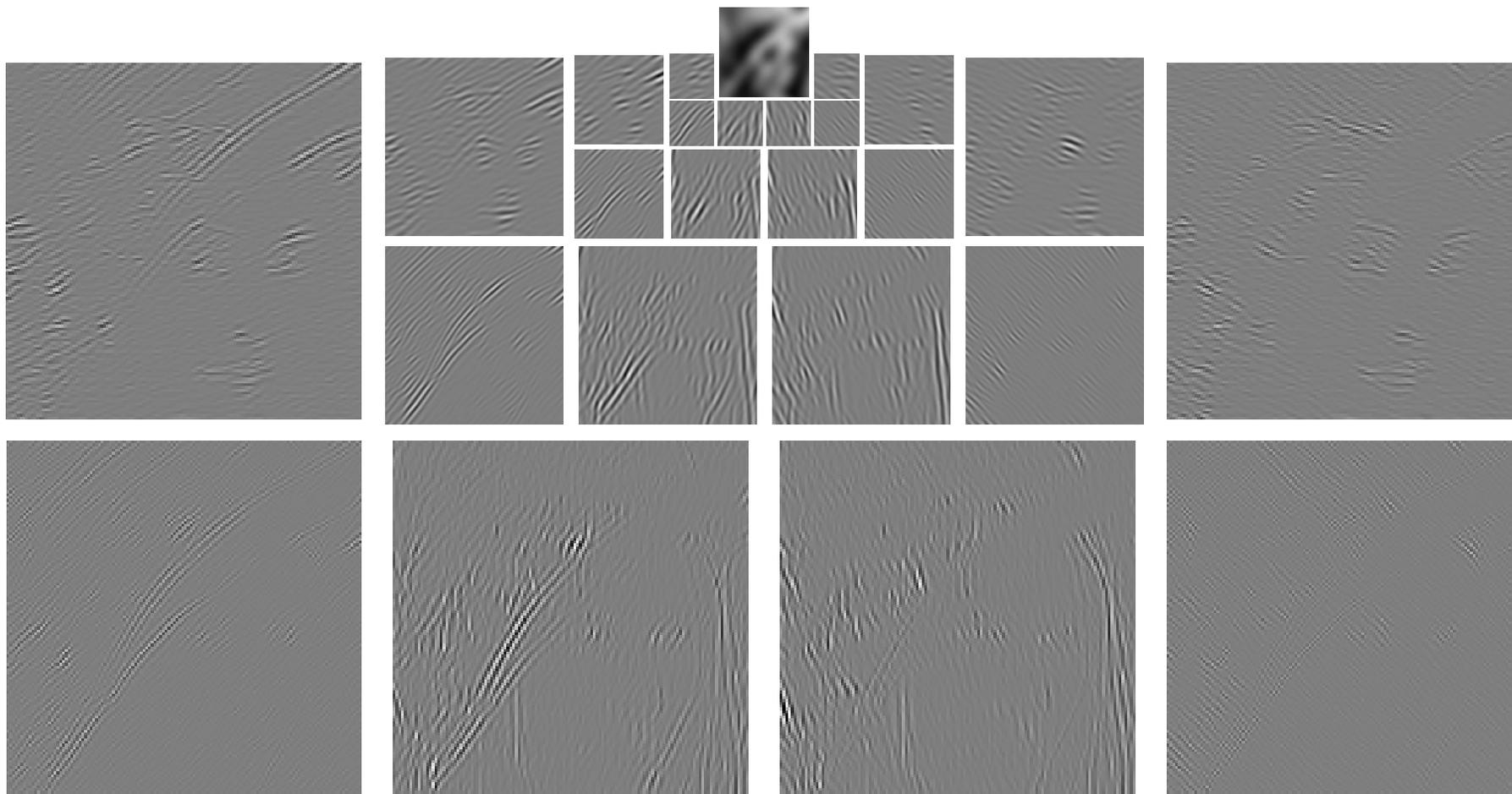


Figure 6: Components from each subband of the reconstructed output image for a 4-level DT-CWT decomposition of *Lenna* (central  $128 \times 128$  portion only).

## 2-D SHIFT INVARIANCE OF DT CWT vs DWT

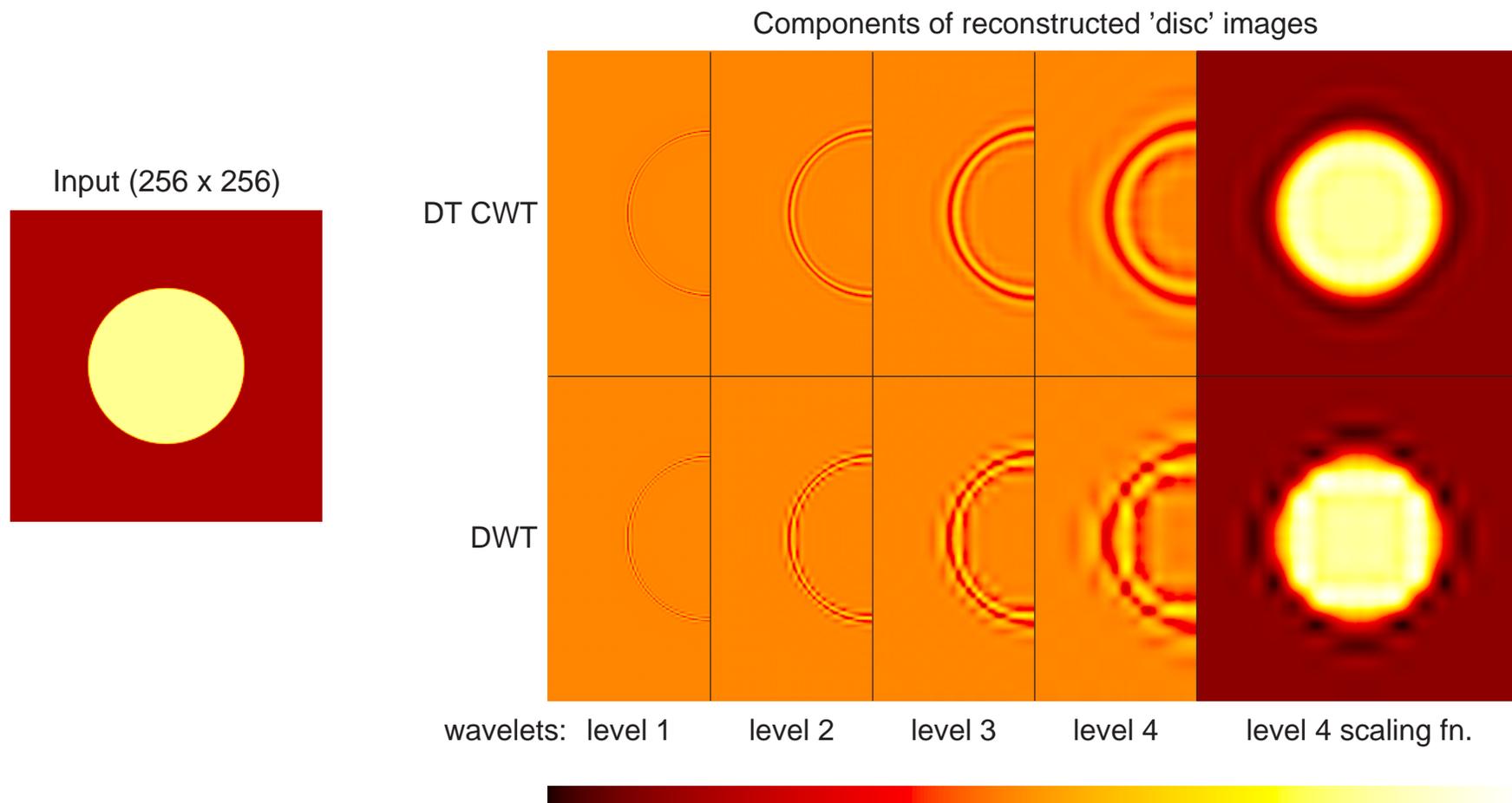


Figure 7: Wavelet and scaling function components at levels 1 to 4 of an image of a light circular disc on a dark background, using the 2-D DT CWT (upper row) and 2-D DWT (lower row). Only half of each wavelet image is shown in order to save space.

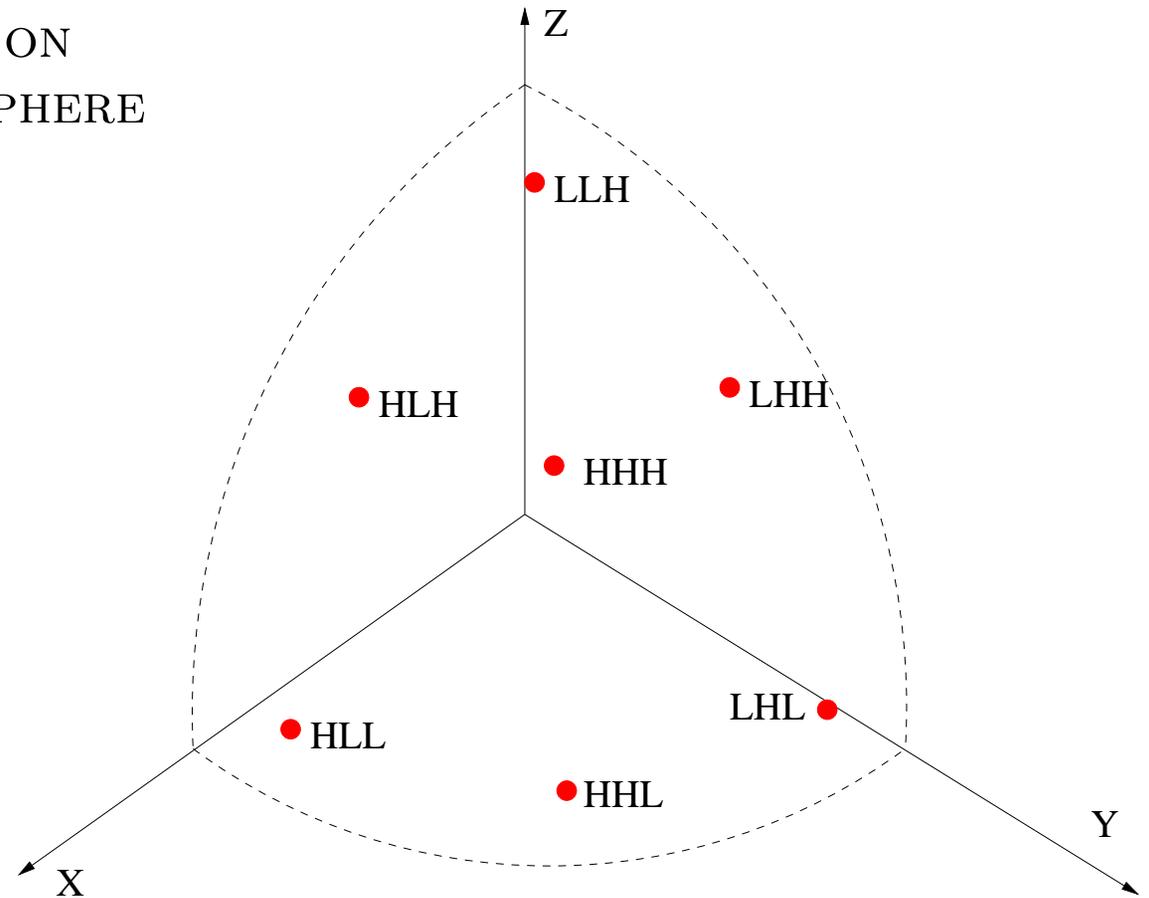
## THE DT CWT IN 3-D

When the DT CWT is applied to 3-D signals (eg medical MRI or CT datasets), it has the following features:

- It is performed separably, with 2 trees used for the rows, 2 trees for the columns and 2 trees for the slices of the 3-D dataset – yielding an **Octal-Tree** structure (8:1 redundancy).
- The 8 octal-tree components of each coefficient are combined by simple sum and difference operations to yield a **quad of complex coefficients**. These are part of 4 separate subbands in adjacent octants of the 3-D spectrum.
- This produces **28 directionally selective subbands** ( $4 \times 8 - 4$ ) at each level of the 3-D DT CWT. The subband basis functions are now **planar waves** of the form  $e^{j(\omega_1 x + \omega_2 y + \omega_3 z)}$ , modulated by a 3-D Gaussian envelope.
- Each subband responds to approximately flat surfaces of a particular orientation. There are 7 orientations on each quadrant of a hemisphere.

### 3D SUBBAND ORIENTATIONS ON ONE QUADRANT OF A HEMISPHERE

3D frequency domain:



3D Gabor-like basis functions:

$$h_{k_1, k_2, k_3}(x, y, z) \simeq e^{-(x^2 + y^2 + z^2)/2\sigma^2} \times e^{j(\omega_{k_1} x + \omega_{k_2} y + \omega_{k_3} z)}$$

These are **28 planar waves** (7 per quadrant of a hemisphere) whose orientation depends on  $\omega_{k_1} \in \{\omega_L, \omega_H\}$  and  $\omega_{k_2}, \omega_{k_3} \in \{\pm\omega_L, \pm\omega_H\}$ , where  $\omega_H \simeq 3\omega_L$ .

## APPLICATIONS OF THE DT CWT

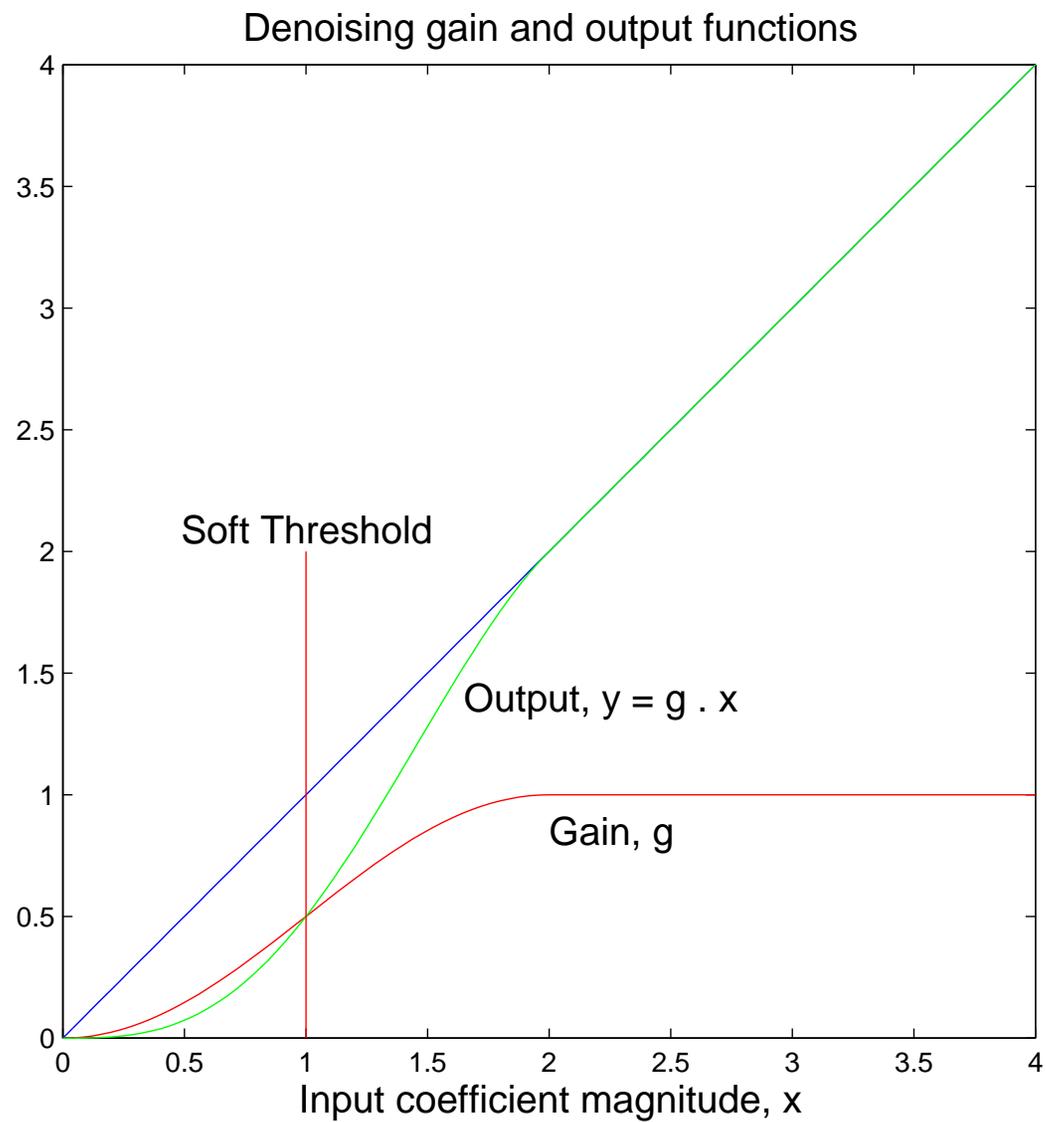
- **Motion estimation** [Magarey 98] and **compensation**
- **Registration** [Kingsbury 02]
- **Denoising** [Choi 00, Miller 06] and **deconvolution** [Jalobeanu 00, De Rivaz 01, J Ng 07]
- **Texture analysis** [Hatipoglu 99] and **synthesis** [De Rivaz 00]
- **Segmentation** [De Rivaz 00, Shaffrey 02]
- **Classification** [Romberg 00] and **image retrieval** [Kam & T T Ng 00, Shaffrey 03]
- **Watermarking of images** [Loo 00] and **video** [Earl 03]
- **Compression / Coding** [Reeves 03]
- **Seismic analysis** [van Spaendonck & Fernandes 02, Miller 05]
- **Diffusion Tensor MRI visualisation** [Zymnis 04]
- **Object matching & recognition** [Anderson & Fauqueur 06]

## DE-NOISING – METHOD:

- Transform the noisy input image to **compress the image energy** into as few coefs as possible, leaving the noise well distributed.
- Suppress lower energy coefs (mainly noise).
- Inverse transform to recover de-noised image.

## WHAT IS THE OPTIMUM TRANSFORM ?

- **DWT** is better than **DCT** or **DFT** for compressing image energy.
- But DWT is **shift dependent** – Is a coef small because there is no signal energy at that scale and location, **or** because it is sampled near a zero-crossing in the wavelet response?
- The **undecimated DWT** can solve this problem but at **significant cost** – redundancy (and computation) is increased by  $3M : 1$ , where  $M$  is no. of DWT levels.
- The **DT CWT** has only  $4 : 1$  redundancy, is directionally selective, and works well.



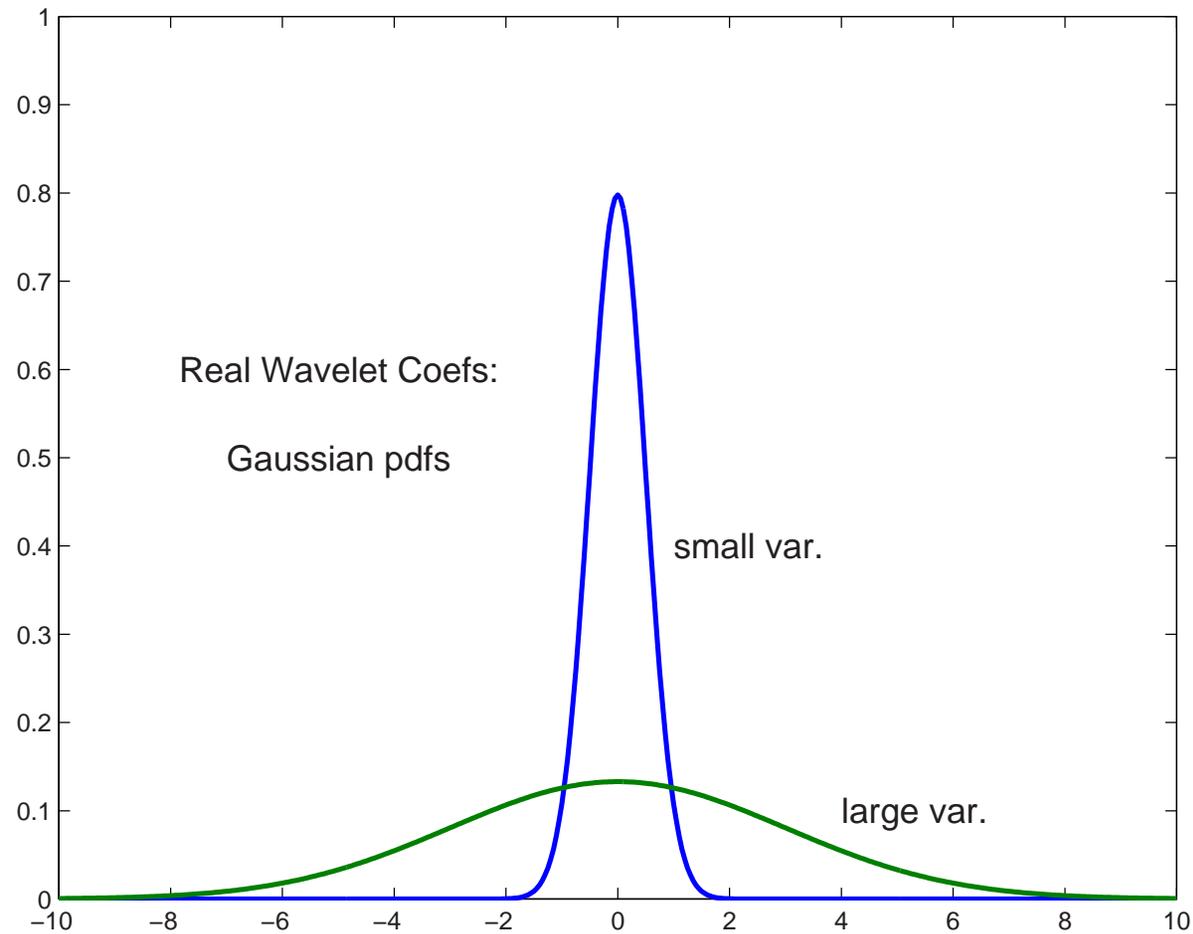


Figure 8: Probability density functions (pdfs) of small and large variance Gaussian distributions, typical for modelling **real and imaginary parts** of complex wavelet coefficients.

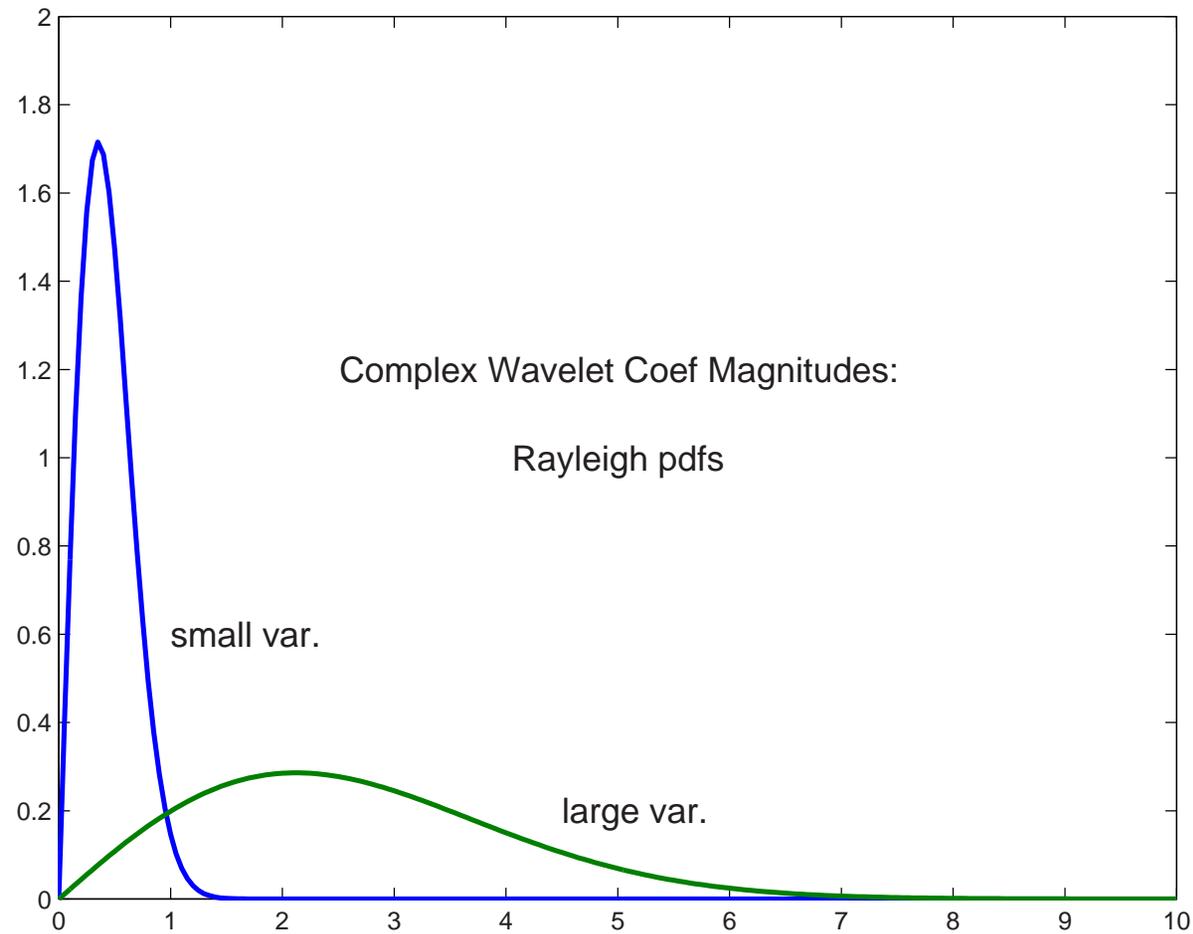


Figure 9: Probability density functions (pdfs) of small and large variance Rayleigh distributions, typical for modelling **magnitudes** of complex wavelet coefficients.

## IMAGE DENOISING WITH DIFFERENT WAVELET TRANSFORMS - LENA

AWGN  
SNR =  
3.0 dB



Real DWT  
SNR =  
11.67 dB



Undec. WT  
SNR =  
12.82 dB



DT CWT  
SNR =  
12.99 dB



## IMAGE DENOISING WITH DIFFERENT WAVELET TRANSFORMS - PEPPERS

AWGN  
SNR =  
3.0 dB



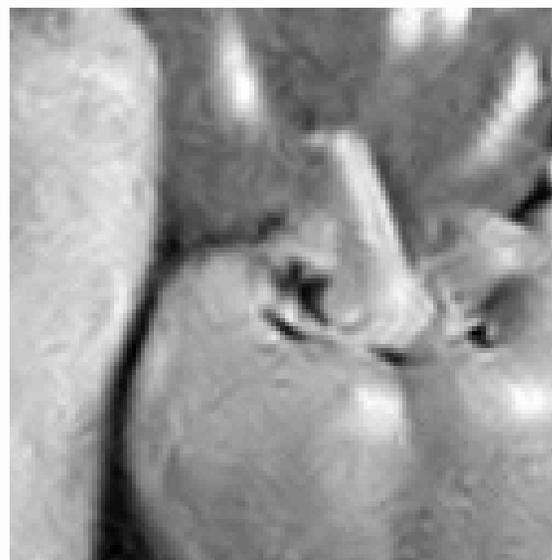
Real DWT  
SNR =  
12.24 dB



Undec. WT  
SNR =  
13.45 dB



DT CWT  
SNR =  
13.51 dB



## HEIRARCHICAL DENOISING WITH GAUSSIAN SCALE MIXTURES (GSMS)

Non-heir.  
DT CWT  
SNR =  
12.99 dB



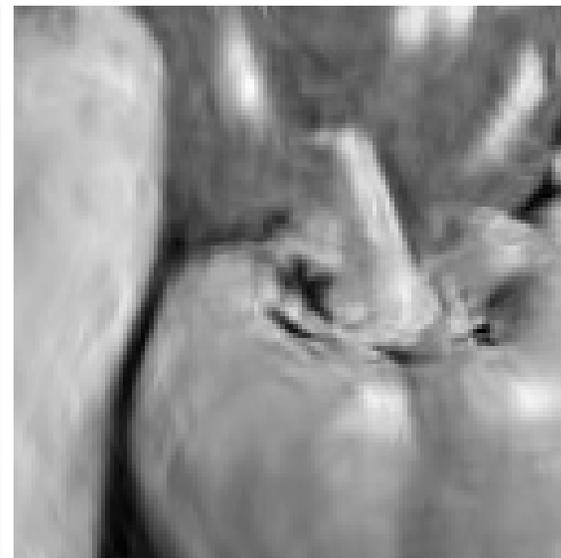
Heirarchical  
DT CWT  
SNR =  
13.51 dB



Non-heir.  
DT CWT  
SNR =  
13.51 dB



Heirarchical  
DT CWT  
SNR =  
13.85 dB



## DENOISING A 3-D DATASET

e.g. Medical 3-D MRI or helical CT scans.

### Method:

- Perform 3-D DT CWT on the dataset.
- Attenuate smaller coefficients, based on their magnitudes, as for 2-D denoising. (Heirarchical methods are also quite feasible.)
- Perform inverse 3-D DT CWT to recover the denoised dataset.

A Matlab example shows denoising of an ellipsoidal surface, buried in Gaussian white noise.

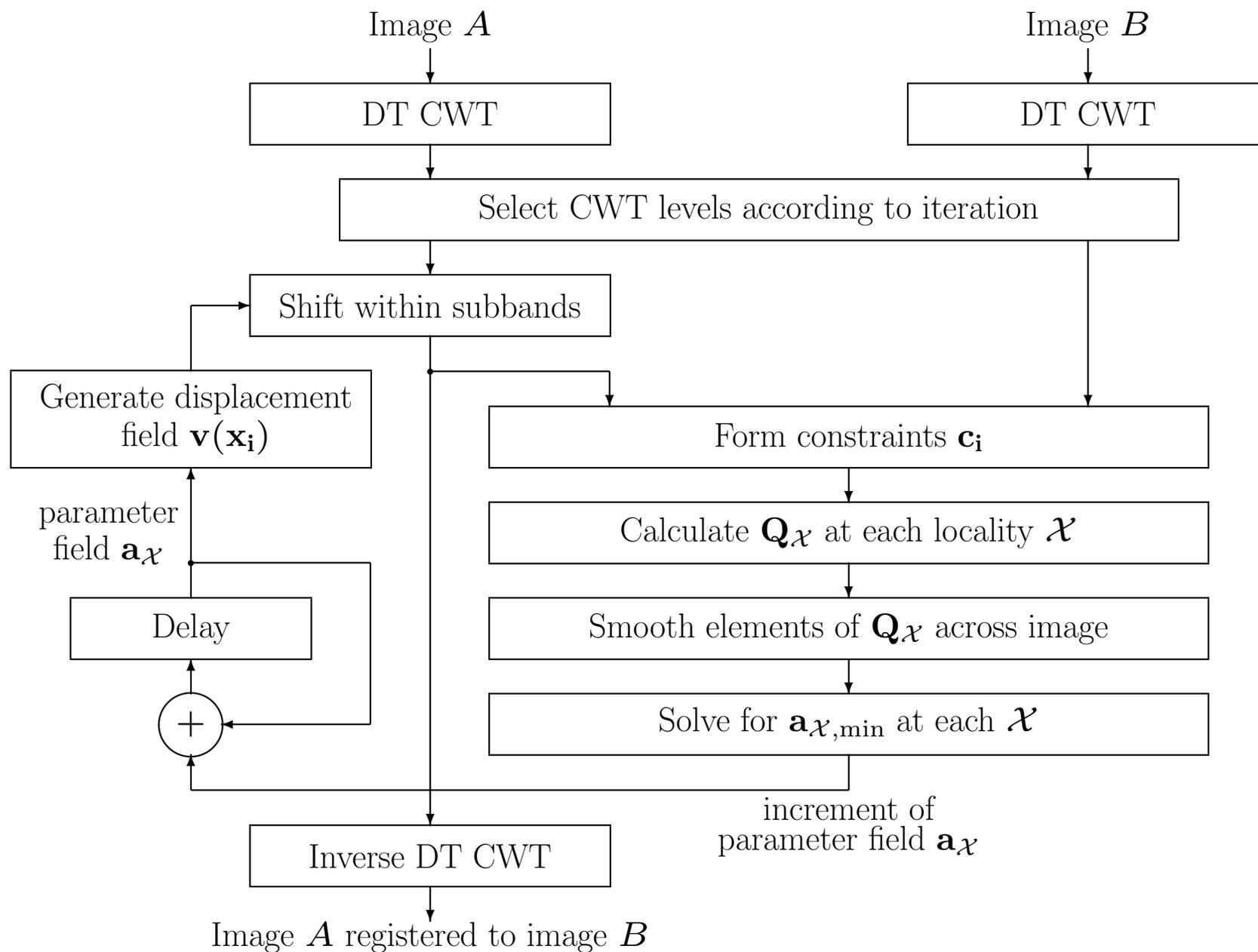
## IMAGE REGISTRATION

### KEY FEATURES OF ROBUST REGISTRATION ALGORITHMS

- Edge-based methods are more robust than point-based ones.
- Must be automatic (no human picking of correspondence points) in order to achieve sub-pixel accuracy in noise.
- Bandlimited multiscale (wavelet) methods will allow spatially adaptive denoising.
- Phase-based bandpass methods can give rapid convergence and immunity to illumination changes between images.
- Displacement field should be smooth, so use of a wide-area parametric (affine) model is preferable to local translation-only models.

## SELECTED METHOD

- Dual-tree Complex Wavelet Transform (DT CWT):
  - provides complex coefficients whose phase shift depends approximately linearly with displacement;
  - allows each subband of coefficients to be interpolated independently of other subbands (because of shift invariance).
  
- Parametric model of displacement field, whose solution is based on local edge-based motion constraints (Hemmendorf et al., IEEE Trans Medical Imaging, Dec 2002):
  - derives straight-line constraints from directional subbands of DT CWT;
  - solves for model parameters which minimise constraint error energy over multiple directions and scales.



## BASIC LINEAR FLOW MODEL

Key Assumption for local translation model:

- Time derivative of the phase  $\theta$  of each complex wavelet coefficient depends **approximately linearly** on the local velocity vector  $\mathbf{v}$ .

This can be expressed as a flow equation in time and spatial derivatives:

$$\frac{\partial \theta}{\partial t} = \nabla_{\mathbf{x}} \theta \cdot \mathbf{v}$$

We can rearrange this to be in the form:

$$\nabla_{\mathbf{x}} \theta \cdot \mathbf{v} - \frac{\partial \theta}{\partial t} = 0$$

or

$$\begin{bmatrix} \nabla_{\mathbf{x}} \theta \\ -\frac{\partial \theta}{\partial t} \end{bmatrix}^T \tilde{\mathbf{v}} = 0 \quad \text{where} \quad \tilde{\mathbf{v}} = \begin{bmatrix} \mathbf{v} \\ 1 \end{bmatrix}$$

## PARAMETRIC MODEL: CONSTRAINT EQUATIONS

Let the displacement vector at the  $i^{\text{th}}$  location  $\mathbf{x}_i$  be  $\mathbf{v}(\mathbf{x}_i)$ ; and let  $\tilde{\mathbf{v}}_i = \begin{bmatrix} \mathbf{v}(\mathbf{x}_i) \\ 1 \end{bmatrix}$ .

A straight-line constraint on  $\mathbf{v}(\mathbf{x}_i)$  can be written

$$\mathbf{c}_i^T \tilde{\mathbf{v}}_i = 0 \quad \text{or} \quad c_{1,i}v_{1,i} + c_{2,i}v_{2,i} + c_{3,i} = 0$$

For a phase-based system in which wavelet coefficients at  $\mathbf{x}_i$  in images  $A$  and  $B$  have phases  $\theta_A$  and  $\theta_B$ , approximate phase linearity means that

$$\mathbf{c}_i = C_i \begin{bmatrix} \nabla_{\mathbf{x}} \theta(\mathbf{x}_i) \\ \theta_B(\mathbf{x}_i) - \theta_A(\mathbf{x}_i) \end{bmatrix}$$

In practise we compute this by averaging finite differences at the centre of a  $2 \times 2 \times 2$  block of coefficients from images  $A$  and  $B$ .

$C_i$  is a constant which does not affect the line defined by the constraint, but which is important later.

## PARAMETERS OF THE MODEL

We can define an affine parametric model for  $\mathbf{v}$  such that

$$\mathbf{v}(\mathbf{x}) = \begin{bmatrix} a_1 \\ a_2 \end{bmatrix} + \begin{bmatrix} a_3 & a_5 \\ a_4 & a_6 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix}$$

or in a more useful form

$$\mathbf{v}(\mathbf{x}) = \begin{bmatrix} 1 & 0 & x_1 & 0 & x_2 & 0 \\ 0 & 1 & 0 & x_1 & 0 & x_2 \end{bmatrix} \cdot \begin{bmatrix} a_1 \\ \vdots \\ a_6 \end{bmatrix} = \mathbf{K}(\mathbf{x}) \cdot \mathbf{a}$$

Affine models can synthesise translation, rotation, constant zoom, and shear.

A quadratic model, which allows for linearly changing zoom (approx perspective), requires up to 6 additional parameters and columns in  $\mathbf{K}$  of the form

$$\begin{bmatrix} \dots & x_1x_2 & 0 & x_1^2 & 0 & x_2^2 & 0 \\ \dots & 0 & x_1x_2 & 0 & x_1^2 & 0 & x_2^2 \end{bmatrix}$$

## SOLVING FOR THE MODEL PARAMETERS

Let  $\tilde{\mathbf{K}}_i = \begin{bmatrix} \mathbf{K}(\mathbf{x}_i) & \mathbf{0} \\ \mathbf{0} & 1 \end{bmatrix}$  and  $\tilde{\mathbf{a}} = \begin{bmatrix} \mathbf{a} \\ 1 \end{bmatrix}$  so that  $\tilde{\mathbf{v}}_i = \tilde{\mathbf{K}}_i \tilde{\mathbf{a}}$ .

Ideally for a given image locality  $\mathcal{X}$ , we wish to find the parametric vector  $\tilde{\mathbf{a}}$  such that

$$\mathbf{c}_i^T \tilde{\mathbf{v}}_i = 0 \quad \text{when} \quad \tilde{\mathbf{v}}_i = \tilde{\mathbf{K}}_i \tilde{\mathbf{a}} \quad \text{for all } i \text{ such that } \mathbf{x}_i \in \mathcal{X}.$$

In practise this is an overdetermined set of equations, so we find the LMS solution, the value of  $\mathbf{a}$  which minimises the squared error

$$\begin{aligned} \mathcal{E}_{\mathcal{X}} &= \sum_{i \in \mathcal{X}} \|\mathbf{c}_i^T \tilde{\mathbf{v}}_i\|^2 = \sum_{i \in \mathcal{X}} \|\mathbf{c}_i^T \tilde{\mathbf{K}}_i \tilde{\mathbf{a}}\|^2 = \sum_{i \in \mathcal{X}} (\tilde{\mathbf{a}}^T \tilde{\mathbf{K}}_i^T \mathbf{c}_i)(\mathbf{c}_i^T \tilde{\mathbf{K}}_i \tilde{\mathbf{a}}) \\ &= \tilde{\mathbf{a}}^T \tilde{\mathbf{Q}}_{\mathcal{X}} \tilde{\mathbf{a}} \quad \text{where} \quad \tilde{\mathbf{Q}}_{\mathcal{X}} = \sum_{i \in \mathcal{X}} (\tilde{\mathbf{K}}_i^T \mathbf{c}_i \mathbf{c}_i^T \tilde{\mathbf{K}}_i) \end{aligned}$$

## SOLVING FOR THE MODEL PARAMETERS (CONT.)

Since  $\tilde{\mathbf{a}} = \begin{bmatrix} \mathbf{a} \\ 1 \end{bmatrix}$  and  $\tilde{\mathbf{Q}}_{\mathcal{X}}$  is symmetric, we define  $\tilde{\mathbf{Q}}_{\mathcal{X}} = \begin{bmatrix} \mathbf{Q} & \mathbf{q} \\ \mathbf{q}^T & q_0 \end{bmatrix}_{\mathcal{X}}$  so that

$$\mathcal{E}_{\mathcal{X}} = \tilde{\mathbf{a}}^T \tilde{\mathbf{Q}}_{\mathcal{X}} \tilde{\mathbf{a}} = \mathbf{a}^T \mathbf{Q} \mathbf{a} + 2 \mathbf{a}^T \mathbf{q} + q_0$$

$\mathcal{E}_{\mathcal{X}}$  is minimised when  $\nabla_{\mathbf{a}} \mathcal{E}_{\mathcal{X}} = 2 \mathbf{Q} \mathbf{a} + 2 \mathbf{q} = \mathbf{0}$ , so  $\mathbf{a}_{\mathcal{X},\min} = -\mathbf{Q}^{-1} \mathbf{q}$ .

The choice of locality  $\mathcal{X}$  will depend on application:

- If it is expected that the affine (or quadratic) model will apply accurately to the whole image, then  $\mathcal{X}$  can be the whole image and maximum robustness will be achieved.
- If not, then  $\mathcal{X}$  should be a smaller region, chosen to optimise the tradeoff between robustness and model accuracy. A good way to produce a smooth field is to make  $\mathcal{X}$  fairly small (e.g. a  $32 \times 32$  pel region) and then to apply a smoothing filter across all the  $\tilde{\mathbf{Q}}_{\mathcal{X}}$  matrices, element by element, before solving for  $\mathbf{a}_{\mathcal{X},\min}$  in each region.

## CONSTRAINT WEIGHTING FACTORS

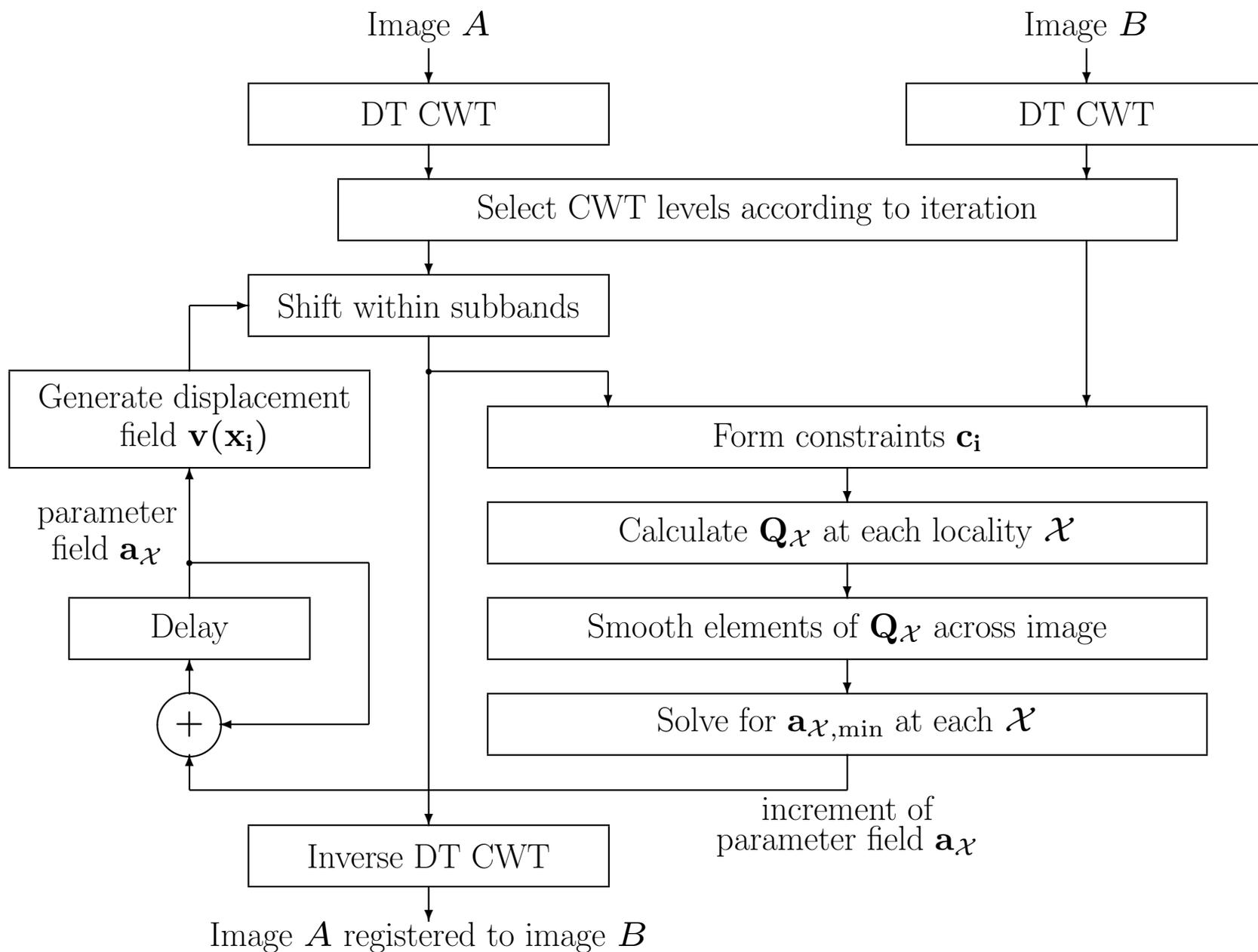
Returning to the equation for the constraint vectors,  $\mathbf{c}_i = C_i \begin{bmatrix} \nabla_{\mathbf{x}} \theta(\mathbf{x}_i) \\ \theta_B(\mathbf{x}_i) - \theta_A(\mathbf{x}_i) \end{bmatrix}$ ,

the constant gain parameter  $C_i$  will determine how much weight is given to each constraint in  $\tilde{\mathbf{Q}}_{\mathcal{X}} = \sum_{i \in \mathcal{X}} (\tilde{\mathbf{K}}_i^T \mathbf{c}_i \mathbf{c}_i^T \tilde{\mathbf{K}}_i)$ .

Hemmendorf proposes some quite complicated heuristics for computing  $C_i$ , but for the DT CWT, we find the following works well:

$$C_i = \frac{|d_{AB}|^2}{4 \sum_{k=1}^4 |u_k|^3 + |v_k|^3} \quad \text{where} \quad d_{AB} = \sum_{k=1}^4 u_k^* v_k$$

and  $\begin{bmatrix} u_1 & u_2 \\ u_3 & u_4 \end{bmatrix}$  and  $\begin{bmatrix} v_1 & v_2 \\ v_3 & v_4 \end{bmatrix}$  are  $2 \times 2$  blocks of wavelet coefficients centred on  $\mathbf{x}_i$  in images  $A$  and  $B$  respectively.

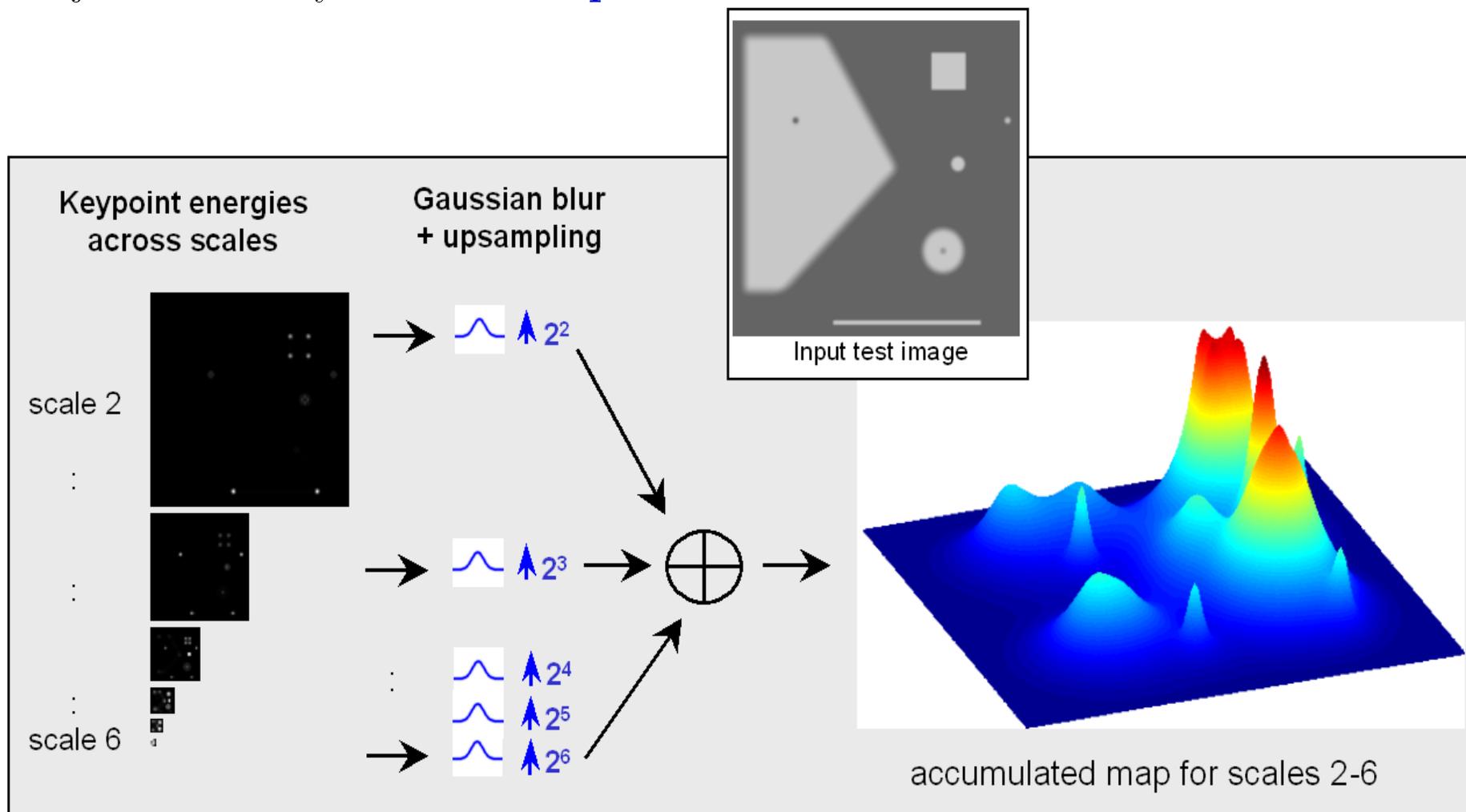


## DEMONSTRATION OF REGISTRATION AND IMAGE FUSION

- House on a hillside, viewed on a video camera with telephoto lens through air with significant heat turbulence (due to a hot runway).
- **Aim:** to recover the best still image from the jittery video sequence of 75 frames.
- Video sequence is courtesy of Don Fraser, Australian Defence Forces Academy, Canberra.
- **Fusion:** based on max of each wavelet coefficient magnitude across the 75 frames, combined with the mean of each coefficient's phase.

## MULTI-SCALE KEYPOINT DETECTION USING ACCUMULATED MAPS

Subject of work by **Julien Fauqueur**.



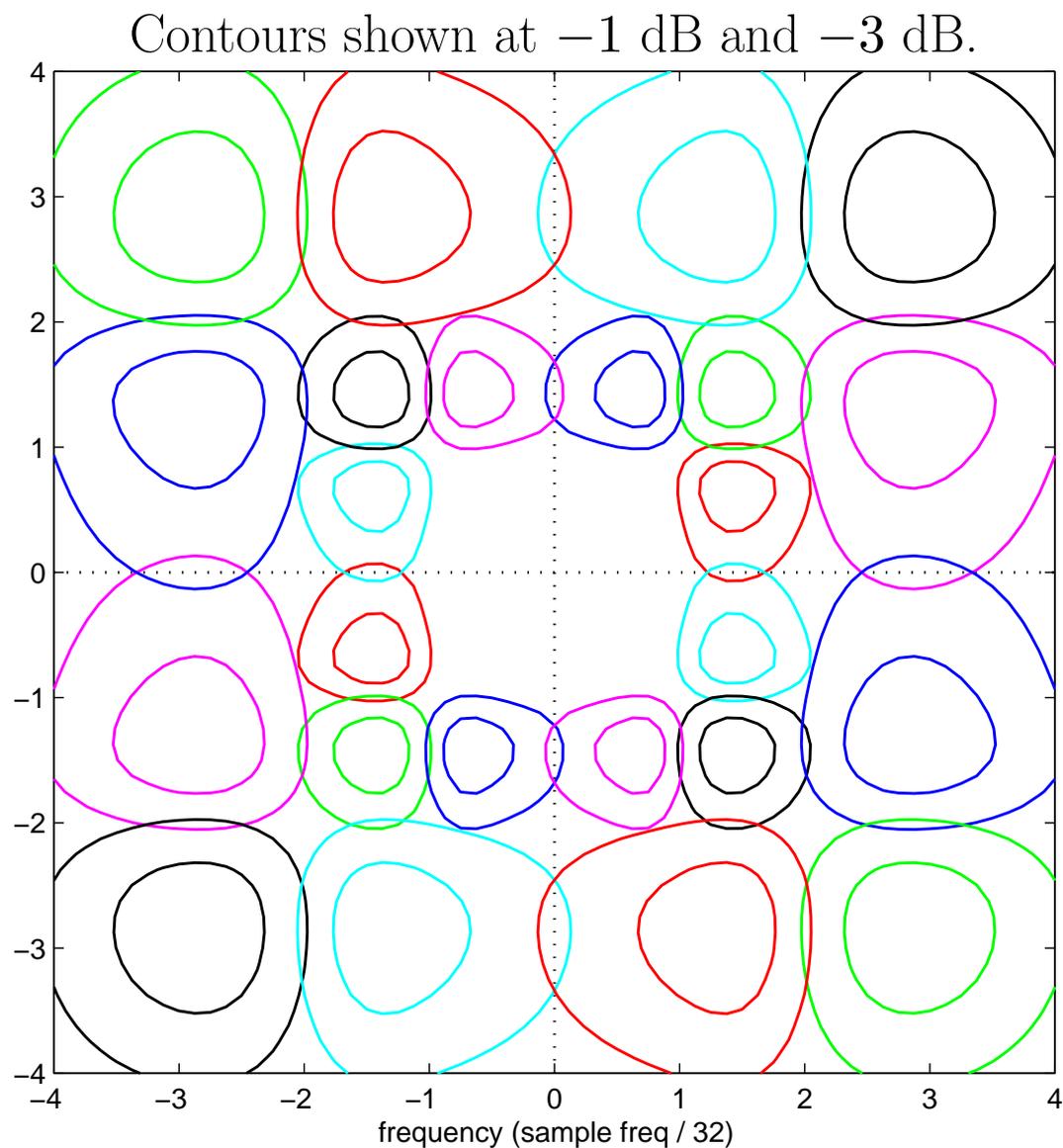
## ROTATION-INVARIANT LOCAL FEATURE MATCHING

### Aims:

- To derive a **local feature descriptor** for the region around a detected keypoint, so that keypoints from similar objects may be **matched reliably**.
- Matching must be performed in a **rotationally invariant** way if all rotations of an object are to be matched correctly.
- The feature descriptor must have **sufficient complexity** to give good detection reliability and low false-alarm rates.
- The feature descriptor must be **simple enough** to allow rapid pairwise comparisons of keypoints.
- Raw DTCWT coefficients provide multi-resolution local feature descriptors, but they are tied closely to a **rectangular sampling** system (as are most other multi-resolution decompositions).

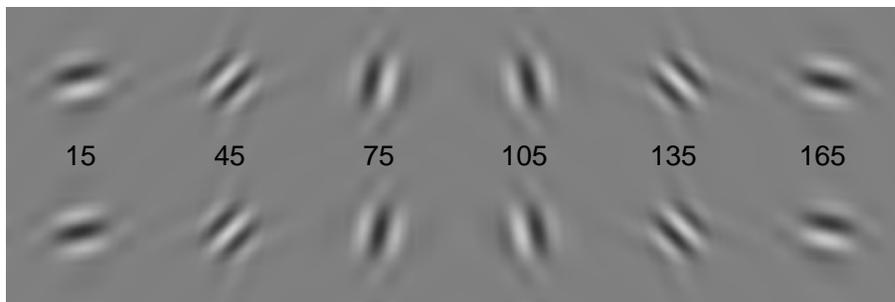
Hence we first need **better rotational symmetry** for the DTCWT.

## FREQUENCY RESPONSES OF 2-D Q-SHIFT FILTERS AT LEVELS 3 AND 4



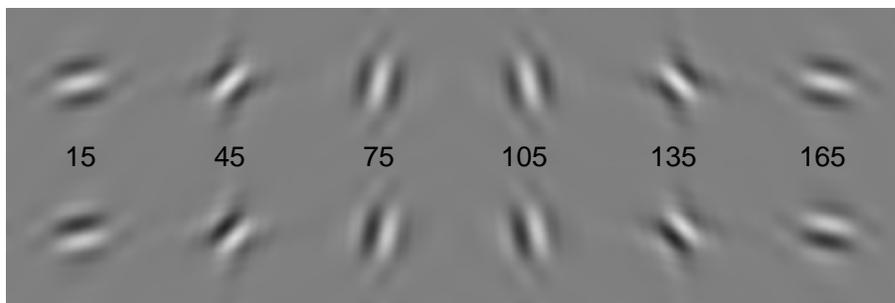
# MODIFICATION OF 45° AND 135° SUBBAND RESPONSES FOR IMPROVED ROTATIONAL SYMMETRY (SHOWN AT LEVEL 4).

(a) Dual-Tree Complex Wavelets: Real Part



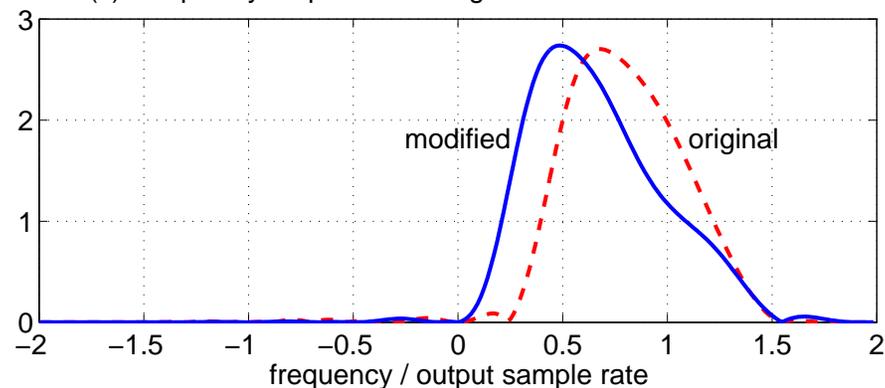
Imaginary Part

(b) Modified Complex Wavelets: Real Part



Imaginary Part

(c) Frequency responses of original and modified 1-D filters

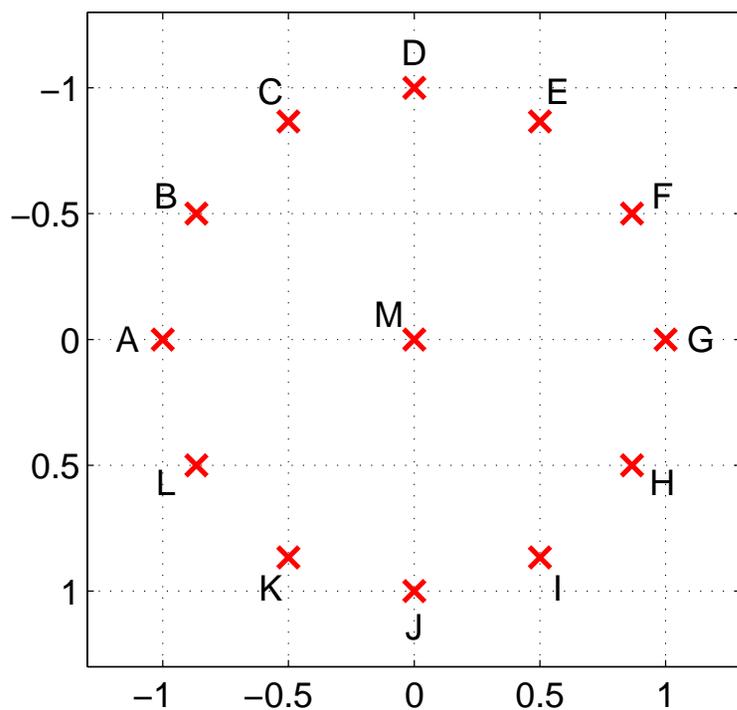


- (a) Original 2-D impulse responses;
- (b) 2-D responses, modified to have lower centre frequencies (reduced by  $1/\sqrt{1.8}$ ) in the 45° and 135° subbands, and even / odd symmetric real / imaginary parts;
- (c) Original and modified 1-D filters.

**Better rotational symmetry** is achieved,  
 but **we have lost Perfect Reconstruction**.

## 13-POINT CIRCULAR PATTERN FOR SAMPLING DTCWT COEFS AT EACH KEYPOINT LOCATION

M is a precise keypoint location, obtained from the keypoint detector.

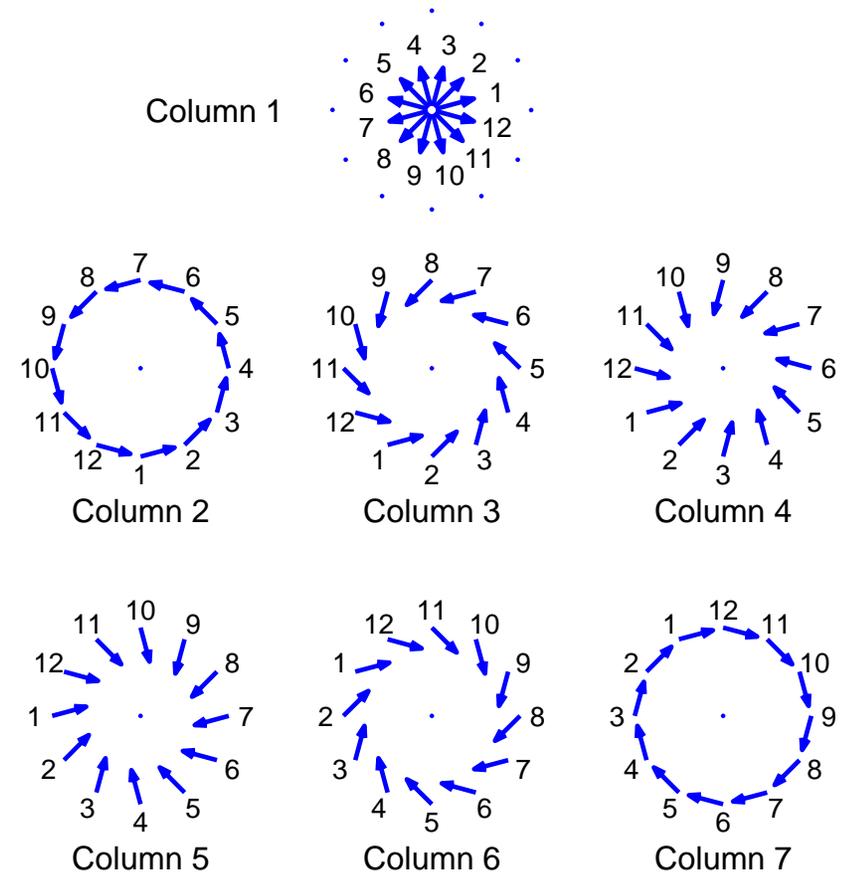


**Bandpass interpolation** calculates the required samples and can be performed on each subband independently because of the shift-invariance of the transform:

1. Shift by  $\{-\omega_1, -\omega_2\}$  down to zero frequency (i.e. multiply by  $e^{-j(\omega_1 x_1 + \omega_2 x_2)}$  at each point  $\{x_1, x_2\}$ );
2. Lowpass interpolate to each new point (spline / bi-cubic / bi-linear);
3. Shift up by  $\{\omega_1, \omega_2\}$  (multiply by  $e^{j(\omega_1 y_1 + \omega_2 y_2)}$  at each new point  $\{y_1, y_2\}$ ).

## FORM THE POLAR MATCHING MATRIX $P$

$$P = \begin{bmatrix} m_1 & j_1 & k_1 & l_1 & a_1 & b_1 & c_1 \\ m_2 & i_2 & j_2 & k_2 & l_2 & a_2 & b_2 \\ m_3 & h_3 & i_3 & j_3 & k_3 & l_3 & a_3 \\ m_4 & g_4 & h_4 & i_4 & j_4 & k_4 & l_4 \\ m_5 & f_5 & g_5 & h_5 & i_5 & j_5 & k_5 \\ m_6 & e_6 & f_6 & g_6 & h_6 & i_6 & j_6 \\ m_1^* & d_1^* & e_1^* & f_1^* & g_1^* & h_1^* & i_1^* \\ m_2^* & c_2^* & d_2^* & e_2^* & f_2^* & g_2^* & h_2^* \\ m_3^* & b_3^* & c_3^* & d_3^* & e_3^* & f_3^* & g_3^* \\ m_4^* & a_4^* & b_4^* & c_4^* & d_4^* & e_4^* & f_4^* \\ m_5^* & l_5^* & a_5^* & b_5^* & c_5^* & d_5^* & e_5^* \\ m_6^* & k_6^* & l_6^* & a_6^* & b_6^* & c_6^* & d_6^* \end{bmatrix}$$



Each column of  $P$  comprises a set of **rotationally symmetric** samples from the 6 subbands and their conjugates (\*), whose orientations are shown by the arrows.

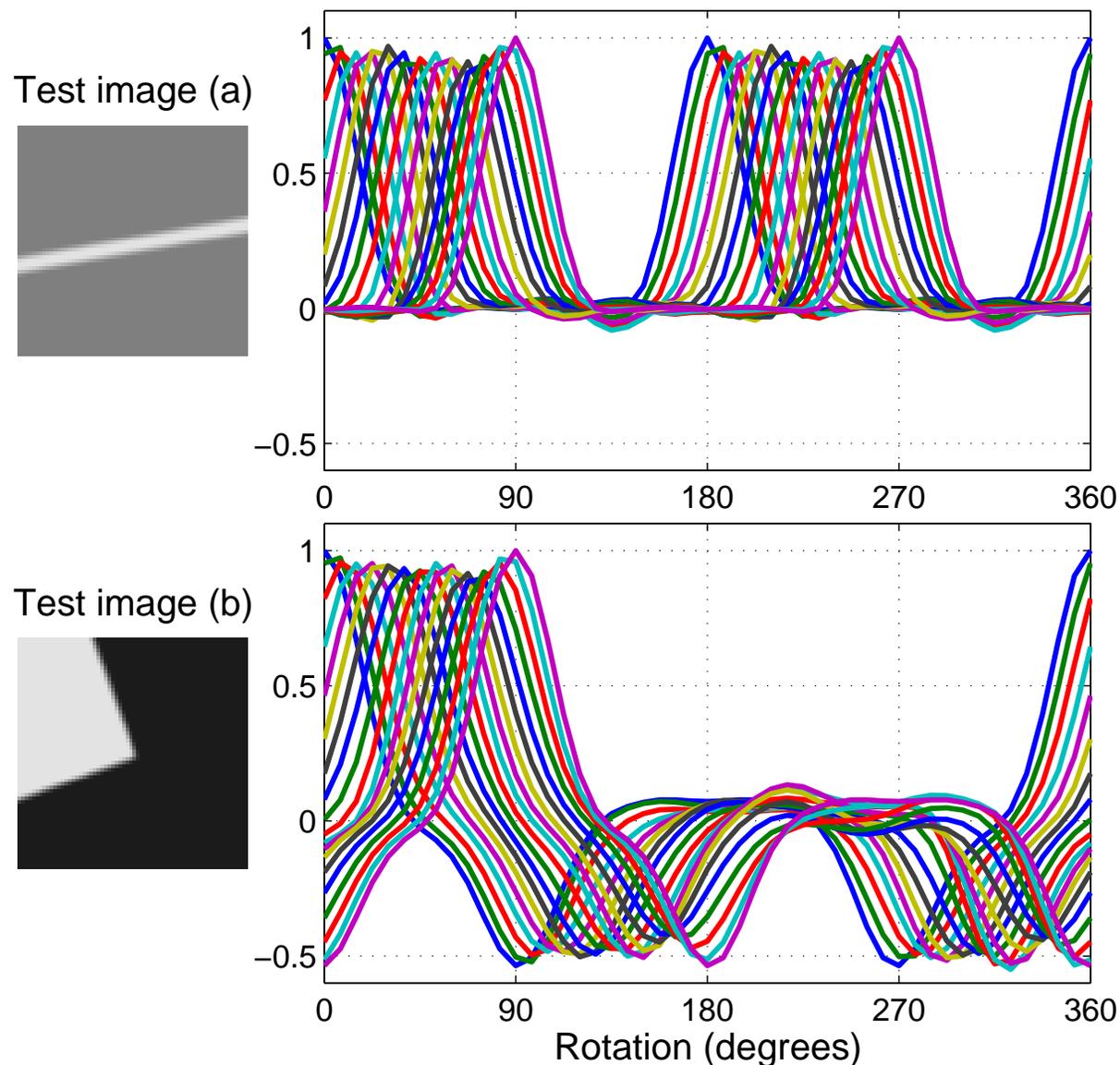
Numbers for each arrow give the row indices in  $P$ .

## EFFICIENT FOURIER-BASED MATCHING

Columns of  $P$  **shift cyclically with rotation** of the object about keypoint  $M$ . Hence we perform correlation matching in the **Fourier** domain, as follows:

- First, take 12-point FFT of each column of  $P_k$  at every keypoint  $k$  to give  $\overline{P}_k$ .
- Then, for each pair of keypoints  $(k, l)$  to be matched:
  - **Multiply**  $\overline{P}_k$  by  $\overline{P}_l^*$  element-by-element to give  $\overline{S}_{k,l}$ .
  - **Accumulate** the 12-point columns of  $\overline{S}_{k,l}$  into a 48-element spectrum vector  $\overline{\mathbf{s}}_{k,l}$  (to give a 4-fold extended frequency range and hence finer correlation steps). Different columns of  $\overline{S}_{k,l}$  are bandpass signals with differing centre frequencies, so optimum interpolation occurs if zero-padding is introduced over the part of the spectrum which is likely to contain least energy in each case.
  - Take the real part of the **inverse FFT** of  $\overline{\mathbf{s}}_{k,l}$  to obtain the 48-point correlation result  $\mathbf{s}_{k,l}$ .
  - The **peak** in  $\mathbf{s}_{k,l}$  gives the **rotation and value** of the best match.
- Extra columns can be added to  $P$  for multiple scales.

## CORRELATION PLOTS FOR TWO SIMPLE IMAGES

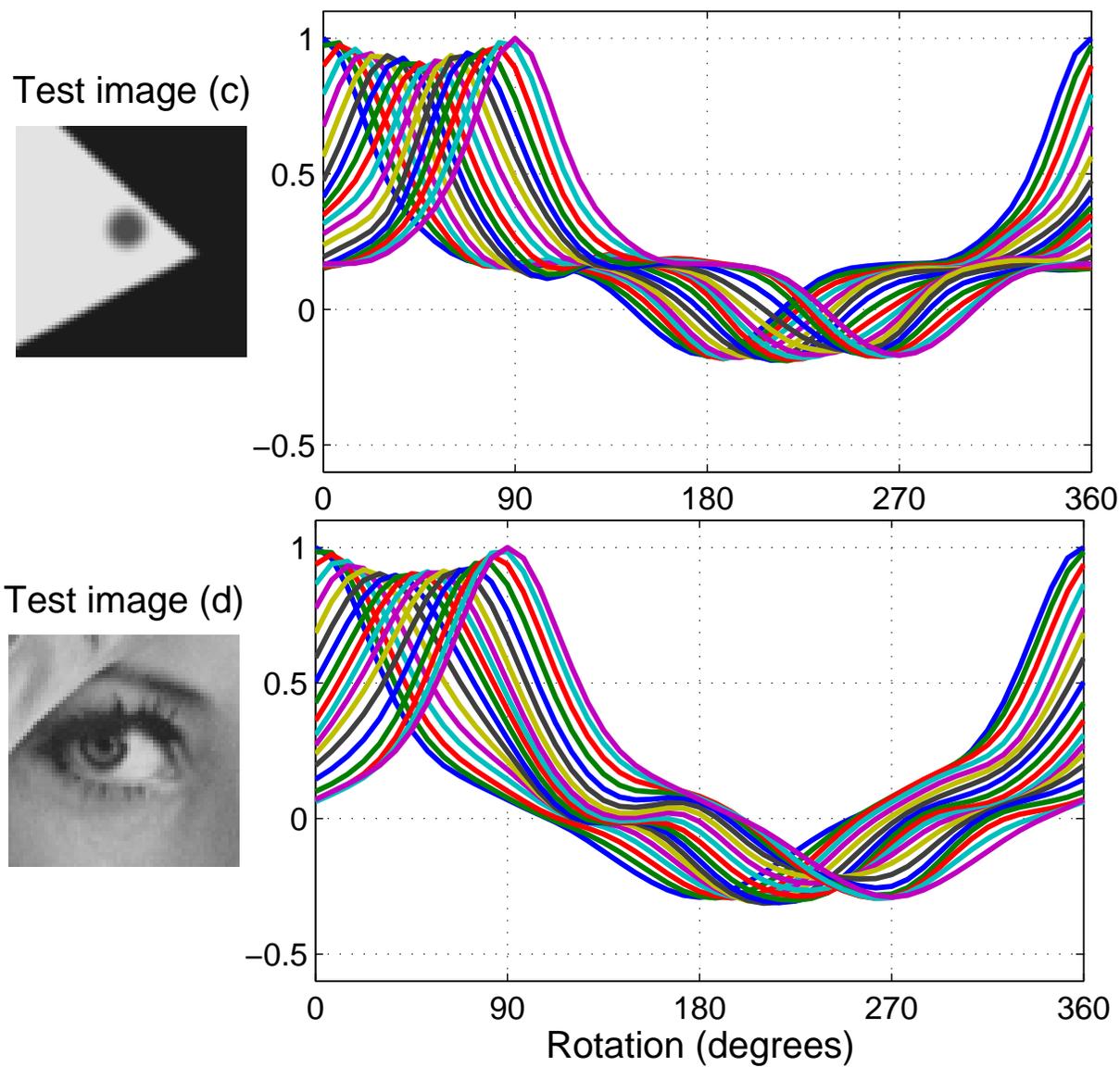


Each set of curves shows the output of the normalised correlator for 48 angles in  $7.5^\circ$  increments, when the test image is rotated in  $5^\circ$  increments from  $0^\circ$  to  $90^\circ$ .

Levels 4 and 5 of the DTCWT were used in an 8-column  $P$  matrix format.

The diameter of the 13-point sampling pattern is half the width of the subimages shown.

## CORRELATION PLOTS FOR MORE COMPLICATED IMAGES



## IMPROVING RESILIENCE TO ERRORS IN KEYPOINT LOCATION AND SCALE

The basic  $P$ -matrix normalised correlation measure is **highly resilient to changes in illumination, contrast and rotation**.

**BUT** it is still rather sensitive to discrepancies in **keypoint location and estimated dominant scale**.

To correct for small errors (typically a few pixels) in keypoint location, we modify the algorithm as follows:

- Measure **derivatives** of  $\bar{P}_k$  with respect to shifts  $\mathbf{x}$  in the sampling circle.
- Using the derivatives, calculate the shift vectors  $\mathbf{x}_i$  which maximise the normalised correlation measures  $\mathbf{s}_{k,l}$  at each of the 48 rotations  $i$  (using LMS methods with approximate adjustments for normalised vectors).
- By regarding the 48-point IFFT as a sparse matrix multiplication, the computation load is only **3 times** that of the basic algorithm.

We propose to do the same for small scale errors using a derivative of  $\bar{P}_k$  wrt scale.

## CONCLUSIONS

The Dual-Tree Complex Wavelet Transform provides **shift invariance** and **orientation selectivity**, in addition to the usual properties of the DWT. We have shown how to apply the DTCWT in the following areas:

- **Denoising** of images and 3D data to achieve performance that equals or exceeds other approaches requiring much more computation.
- **Image Registration** with an efficient multi-resolution iterative algorithm - particularly suited to non-rigid motion.
- **Rotation-invariant local feature matching** at detected keypoints for object detection and keypoint matching applications.

Papers on complex wavelets are available at:

<http://www.eng.cam.ac.uk/~ngk/>

A Matlab DTCWT toolbox is available on request from:

[ngk@eng.cam.ac.uk](mailto:ngk@eng.cam.ac.uk)